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TWO APPROACHES TO  
MIN - MAX FEEDBACK CONTROL AND  
DESIGN OF UNCERTAIN SYSTEMS

by

İbrahim EKSİN

Bogazici University Library



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## A B S T R A C T

In designing control systems with optimal performance parameter variations pose a great problem. In principle, it is possible to attain an ideal performance by sensing the uncertain parameters and using an adaptive controller. But such controllers are usually difficult or expensive to realize and there is a strong motivation for considering simpler controllers.

The simpler control strategy under parameter uncertainty is the min-max controller which operate the system according to one control law that minimizes the maximum variations from the optima corresponding to each parameter. There exist various methods to find the parameters of the min-max controller.

In this study two approaches have been considered to find the min-max controller parameters: iterative or search procedure, and game theoretic approach.

# CHAPTER I

## INTRODUCTION

### 1.1 PROBLEM STATEMENT

Intelligent design of an optimum control system can be carried out only if the designer knows the dynamic characteristics of the plant or process to be controlled. The extent to which system design can proceed in a logical, systematic, and intelligent manner is, to a considerable degree, directly measured by the knowledge of the process dynamics. Thus, the first thing in control system design is to determine the dynamic characteristics of the process to be controlled. Figure 1-1. below illustrates the block diagram of a plant.

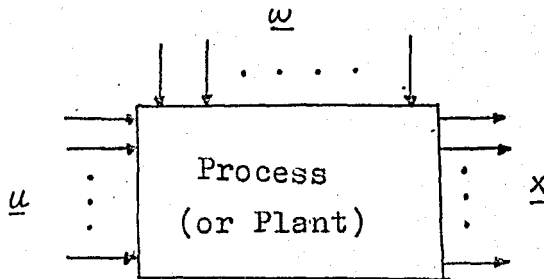


Fig. 1-1. A Multivariable process.

The dynamic characterization of a plant is usually described by a set of differential equations

$$\dot{x}_i = \frac{dx_i(t)}{dt} = f_i [x_1(t), x_2(t), \dots, x_n(t); u_1(t), \dots, u_m(t); w_1(t), \dots, w_s(t); t]$$

$$i = 1, 2, 3, \dots, n$$

In vector-matrix notation, this set of equations may be written as

$$\dot{\underline{X}}(t) = \frac{d\underline{X}(t)}{dt} = \underline{f}[\underline{X}(t), \underline{U}(t), \underline{W}(t), t]$$

In above equations, the plant is assumed to be of n'th order;  $\underline{X}(t)$  is defined as the state vector

$$\underline{X}(t) = [X_1(t), X_2(t), \dots, X_n(t)]^T$$

whose n components are the state variables;  $\underline{U}(t)$  is the m-dimensional control vector

$$\underline{U}(t) = [U_1(t), U_2(t), \dots, U_m(t)]^T$$

whose m components are control signals;  $\underline{W}(t)$  is an s-dimensional vector-valued random function known as the disturbance vector

$$\underline{W}(t) = [W_1(t), W_2(t), \dots, W_s(t)]$$

whose s components represent the random disturbances, and  $\underline{f}$  is a known vector function.

At each moment, the control signals must satisfy the inequalities

$$g_j(U_1, U_2, \dots, U_m) \leq 0 \quad j = 1, 2, \dots, m$$

or in vector notation

$$\underline{g}(\underline{U}) \leq 0 \quad (\text{where } g \text{ is a general function})$$

which reflects the restrictions imposed upon the control system. The control vector  $\underline{U}$  which satisfies the above inequality is referred to as the admissible control vector. For instance, in many practical situations the values of the control signals can

not exceed certain upper bounds because of saturation effects or physical limitations. Under such circumstances, the admissible control signals must satisfy the inequalities

$$|U_j| \leq M_j \quad j = 1, 2, \dots, m$$

Now, considering the design of an optimum control system, if the designer does not know the dynamic characteristics (parameters) of the plant or process exactly, he can proceed in the following two ways:

#### A) System Identification

One tries to estimate the parameters of the system (thus identify the system) by gathering some statistical data about the system and using certain techniques (such as Least Square, Maximum Likelihood, Gradient Method, MAP, etc.) Afterwards, the optimum control law is found out in a conventional manner, with known plant dynamics.

#### B) Min-Max Approach

In this case, the designer considers the system parameters as uncertain over a range. He assumes that the worst will happen; and accordingly, he tries to choose a controller which results in the least drastic effect on the system performance.

In this study the second approach (Min-Max approach) will be used. Just like in parameter estimation there are various methods to solve min-max problem and here in this study the two most common methods will be used to solve it.

The uncertainty in the system parameters may arise because of lack of precise information about a specific system to be controlled or the requirement to determine a fixed

controller suitable for use with an ensemble of systems differing in the values by some group of parameters. In such situations two reasonable design criteria can be immediately conjectured: 1) Minimize the maximum deviation from the optimal behaviour; and 2) Minimize the average deviation from optimal behaviour. Either of these criteria might be applied when the designer is attempting to find a single controller for a number of similar plants or when he is attempting to find a controller for a single changeable plant. The first criterion is, of course, the more meaningful when critical tolerances are present, the second, however, would probably find more "production line" use.

Although the methods which will be used in this thesis are applicable to nonlinear systems, here an uncertain linear autonomous system will be considered which may be modelled by the vector differential equation.

$$\dot{\underline{X}}(t) = \underline{A}(\underline{v})\underline{X}(t) + \underline{B}(\underline{v})\underline{U} \quad \underline{X}(t_0) = \underline{X}_0$$

where  $\underline{X}$  is an  $n \times 1$  state vector,  $A$  is an  $n \times n$  matrix,  $B$  is an  $n \times m$  matrix, and  $U$  is an  $m \times 1$  control vector, and  $v$  is a set of  $p$  time-invariant parameters. The uncertainty is introduced by assuming that the actual value of  $v$  is not known but may be assumed to lie in the compact; i.e., closed and bounded set  $V$ . The performance index to be minimized is

$$J(u, v) = \int_0^{\infty} (\underline{x}^T \underline{Q} \underline{x} + \underline{u}^T \underline{P} \underline{u}) dt$$

where  $P$  is a positive definite  $n \times n$  matrix and  $Q$  is a positive semidefinite  $n \times n$  matrix.

Let  $\mathcal{F}$  be the class of all admissible controls and  $F^0$  be the subset of the set of all admissible controls which are optimal for some parameter  $\underline{v} \in V$ .

Then, find a control  $u^0$  and a parameter  $v^0$  such that

$$|J(u^0, v^0) - J(u, v)| \leq \max |J(u, v^0) - J(u, v)| \quad \text{for all } u \in$$

### 1.2 MIN-MAX PHILOSOPHY

A min-max controller is one which is least sensitive to parameter variations, within an assumed range in terms of the changes in the performance index. We know that an optimal control is evaluated at a particular set of parameters, and the variation in parameters requires a fresh determination of the "control law" to ensure optimality. One solution would be to operate the system according to one control law that minimizes the maximum variations from the respective optima; i.e. min-max controller. The philosophy may be explained with the aid of the figure 1-2.

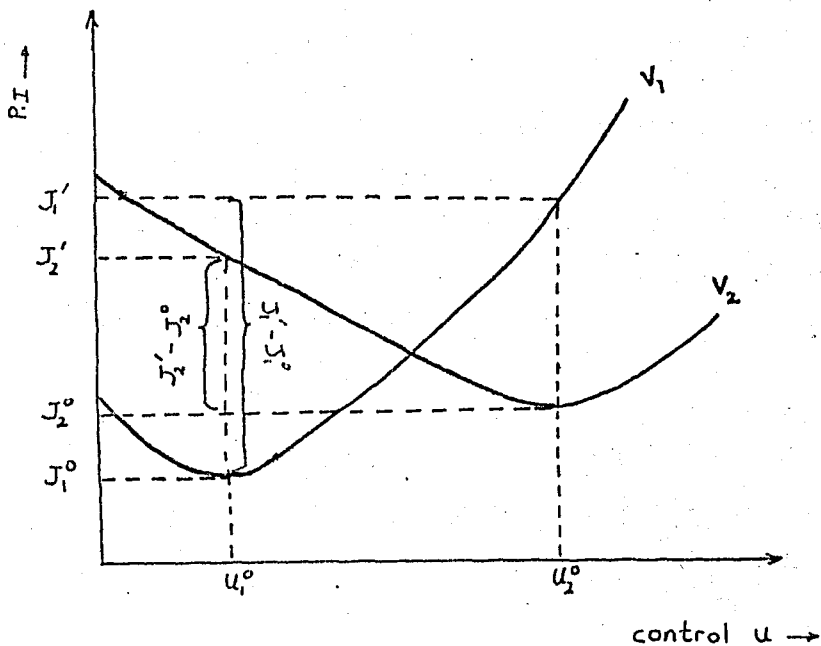


Fig. 1-2. Illustration of min-max concept.

Performance index  $J$  is represented as a function of a single variable  $u$ . But this is adequate for the sake of explanation. Figure 1.2. shows the plots of the cost functionals for two values of parameter  $v_1$  and  $v_2$ . For parameter value  $v_1$ , the optimal control is  $u_1^0$  and the optimal value of the cost functional is  $J_1^0$ . For parameter value  $v_2$ , the optimal control is  $u_2^0$  and the optimal value of the cost functional is  $J_2^0$ . If the system initially has the parametric value  $v_1$ , its control is set at  $u_1^0$  resulting in optimal performance. If the parametric value changes to  $v_2$ , the value of the cost functional would be  $J_2^1$ , the resulting difference from the optimal value being  $(J_2^1 - J_2^0)$ . Similarly, if the control is set at  $u_2^0$  corresponding to the parameter value  $v_2$ , and if the parameter assumes a value  $v_1$ , the resulting perturbation from the optimal value would be  $(J_1^1 - J_1^0)$ . As in Fig 1.2. one has

$$J_1^1 - J_1^0 > J_2^1 - J_2^0 .$$

That is to say  $u_1$  is a better control, since it results in a lesser value of the maximum deviation from the optimum. If  $v_1$  and  $v_2$  are the only parameter values for the system,  $u_1^0$  is clearly the min-max control. The argument can be extended to any number of parameter values.

### 1.3 METHODS OF SOLVING MIN-MAX PROBLEMS

The available methods for solving min-max problems are

- 1) To find an analytic solution to the maximization step, if it exists, and then minimize
- 2) To locate a saddle point if it exists and show that it represents a global solution

- 3) To search or use an iterative procedure
- 4) To use a game theoretic approach.

Experience has shown that the first two methods are seldom useful in practical controller design. For instance, the maximizing solution can not be found in a tractable form, or it is not unique and thus, the minimization step is not possible in the first method. In the second method, a saddle point solution does not exist in general. That is why, in this study to find the solution of the min-max problem the last two methods are considered.

A brief overview of the remaining parts of this study could be summarized as follows:

In chapter 2 a brief summary of some important concepts in game theory and their relation to control problems will be given. In chapters 3 and 4 the two methods, namely the game theoretic approach and the grid-point search methods will be introduced, respectively. In chapter 5 three specific problems which are solved by the two methods will be given. These three specific problems involve from one to three parameter uncertainties. Finally, chapter 6 will be a conclusion and comparison chapter.

In this study the numerical solutions of two famous matrix equations were needed. One of them was Algebraic Matrix Riccati Equation (ARE) and the other was Liapunov Equation. In Appendix A Algebraic Riccati Equation (ARE) solution with an example problem is given and in Appendix B Liapunov Equation solution with an example problem is given. In Appendix C the proof of the min-max theorem is given.

## CHAPTER II

## GAME THEORY

## 2.1 DISCRETE GAMES AND SOME DEFINITIONS

Elementary game theory is concerned with discrete optimization problems involving two players with conflicting interests. In a typical matrix game there are two players,  $u$  and  $v$ , and a selection of strategies,  $u_i$ ,  $i = 1, 2, \dots, m$  and  $v_j$ ,  $j = 1, 2, \dots, n$  for each player.

## DEFINITION:

By a strategy for  $u$  in a matrix game it is meant a decision by  $u$  to play the various rows with a given probability distribution, say to play row one with probability  $P_1$ , to play row two with probability  $P_2$ .

This strategy for  $u$  is formally denoted by the probability vector  $\underline{P} = (P_1, P_2, P_3, \dots, P_m)$ . For example, if the matrix game has two rows and  $u$  tosses a coin to decide which row to play, then this strategy is the probability vector  $\underline{P} = (1/2, 1/2)$ . A strategy which contains a 1 as a component and 0, i.e., where  $u$  decides to play a given column is called PURE STRATEGY; otherwise, it is called MIXED STRATEGY.

For each pair of strategies there is a corresponding pay off,  $J = L_{ij}$  where  $L_{ij}$  are elements of a matrix game.

## DEFINITION:

A game is said to be ZERO-SUM GAME if and only if the pay off function  $(l_1, l_2, \dots, l_n)$  satisfies

$$\sum_{j=1}^n l_j = 0$$

In general, a zero-sum game represents a closed system: everything that one wins must be lost by someone else. Two person zero-sum games are sometimes called strictly competitive games. Player  $u$  attempts to minimize his pay off, while  $v$  attempts to maximize his pay off. This is a "perfect information game", in the sense that each player has all the information above and that each player knows the other's choice of strategies. Now, if  $V$  (the maximizer) plays first he should obviously pick the column with largest minimum since he knows  $u$  will subsequently pick the row with the minimum. Similarly, if  $u$  (the minimizer) plays first he should pick the row with the smallest maximum, since he knows  $v$  will subsequently pick the column with the maximum.

#### Example 2.1

Consider the following matrix game:

		$V$		
		$v_1$	$v_2$	
$U$	$u_1$	$L_{11}=2$	$L_{12}=7$	← row with smaller maximum
	$u_2$	$L_{21}=5$	$L_{22}=9$	
		↑ └ column with larger minimum		

Figure 2-1. A simple discrete game with a saddle point.

The optimal choices for the above game are  $u_1$  and  $v_2$ , with payoff 7, regardless of who plays first, i.e., we have

$$\max_{v_j} \min_{u_i} L_{ij} = 7 = \min_{u_i} \max_{v_j} L_{ij}$$

( $V$  plays first)

( $U$  plays first)

$$L(u_1, v_j) \leq L(u_1, v_2) \leq L(u_i, v_2)$$

The choice  $u_1, v_2$  is called the minimax solution of the game. And the payoff element  $L_{12}$  is called saddle point.

However, the choice is not always so simple; suppose that the value of  $L_{11}$  is changed from 2 to 11.

Example 2-2.

V is maximizing

	$v_1$	$v_2$
$u_1$	$L_{11}=11$	$L_{12}=7$
$u_2$	$L_{21}=5$	$L_{22}=9$

U

U is minimizing

Figure 2-2. A discrete game where order of play makes a difference.

Then we have

$$\begin{array}{l} \max_v \min_u = 7 \quad \min_u \max_v = 9 \\ (V \text{ plays first}) \quad (u \text{ plays first}) \end{array}$$

If  $v$  (the maximizer) plays first, he should pick  $v_2$ , since this is the column with the larger minimum, namely, 7. If  $u$  (the minimizer) plays first, he should pick  $u_2$ , since

this is the row with the smaller maximum, namely, 9. Thus, it makes a difference who plays first, and we say that there is not a "saddle point" (or a minimax solution). This dilemma may be resolved by having each side make a random selection of strategies on each play according to some fixed probability. Since we are considering probabilistic strategies, it is clear that we will deal with mixed strategies.

## 2.2 MIXED STRATEGIES

In mixed strategy the strategy should be chosen at random but the randomization scheme should be chosen using a specific technique. Before going on further, let's give the definition of mixed strategy.

### DEFINITION:

A mixed strategy for a player is a probability distribution on the set of his pure strategies. For simplicity, let's consider a 2 x 2 matrix game, say

$$\underline{L} = \begin{bmatrix} l_{11} & l_{12} \\ l_{21} & l_{22} \end{bmatrix}$$

Suppose player u adopts the strategy  $\underline{P} = (P_1 \ P_2)$  and player v adopts the strategy  $\underline{q} = (q_1 \ q_2)$ . Then U plays row 1 with probability  $p_1$  and V plays column 1 with probability  $q_1$ , and so the entry  $l_{11}$  will occur with probability  $p_1 q_1$ . Hence, the expected winning of U is

Expected payoff =  $E(p, q) = p_1 q_1 l_{11} + p_1 q_2 l_{12} + p_2 q_1 l_{21} + p_2 q_2 l_{22}$

$$= [p_1 \ p_2] \begin{bmatrix} l_{11} & l_{12} \\ l_{21} & l_{22} \end{bmatrix} \begin{bmatrix} q_1 \\ q_2 \end{bmatrix} = \underline{p} \underline{L} \underline{q}^t = \underline{q} \underline{L}^t \underline{p}$$

Player V will fear that U will discover his choice of strategy. If this should happen, then U will certainly choose  $p$  so as to minimize  $E(p, q)$ ; i.e., V's expected gain-floor assuming he uses  $q$ , will be

$$V(q) = \min_{p \in P} \underline{p} \underline{L} \underline{q}^t = \min_{p \in P} \underline{q} \underline{L}^t \underline{p}^t$$

Thus, the minimum will be attained by a pure strategy,  $i$

$$V(q) = \min_j \underline{L}_j \cdot \underline{q}^t \quad (\underline{L}_j \text{ is the } i^{\text{th}} \text{ row of the matrix } \underline{L})$$

Hence, player V should choose  $q$  so as to maximize  $V(q)$

$$V_I = \max_{q \in Q} \min_j \underline{L}_j \cdot \underline{q}^t$$

One should prove that a maximum exists; however, since  $Q$  is compact and the function  $V(q)$  is continuous, a maximum exists. Such a  $q$  is V's max-min strategy.

Similarly, if U chooses  $p$  he will obtain the expected loss-ceiling.

$$V(p) = \max_{q \in Q} \underline{p} \underline{L} \underline{q}^t = \max_{q \in Q} \underline{q} \underline{L}^t \underline{p}^t$$

The maximum will be obtained by a pure strategy,  $j$

$$V(p) = \max_j \underline{p} \underline{L}_j \quad (\underline{L}_j \text{ is the } j^{\text{th}} \text{ column of the matrix } \underline{L})$$

and the player U should choose  $p$  so as to obtain

$$V_{II} = \min_{p \in P} \max_j p \cdot L_j$$

Such a  $p$  is U's min-max strategy.

Thus we obtain the two numbers  $V_I$  and  $V_{II}$ . These two numbers are called the values of the game to V and U, respectively.

### 2.3 THE MIN-MAX THEOREM

It is easily proved that, for any function  $F(x,y)$  defined on any cartesian product  $X \times Y$

$$\max_{x \in X} \min_{y \in Y} F(x,y) \leq \min_{y \in Y} \max_{x \in X} F(x,y)$$

Hence we have  $V_I \leq V_{II}$

It is very natural that V's gain-floor can not exceed U's loss-ceiling.

#### THE MIN-MAX THEOREM :

The min-max and the max-min values are equal in expected value basis, i.e.,

$$V_I = V_{II}$$

Proof : Reference 12

Let's try to illustrate this theorem by the example 2.3

V is maximizer

	$v_1$	$v_2$
$u_1$	$L_{11}=11$	$L_{12}=7$
$u_2$	$L_{21}=5$	$L_{22}=9$

U is minimizer

If V plays a fixed choice while U uses a random choice, the expected payoffs for various probability mixes of  $U_1$ , and  $U_2$  is shown in the figure 2.3.

Similarly, if U plays a fixed choice while V uses a random choice, the expected payoffs for various probability mixes of  $v_1$  and  $v_2$  is shown in the figure 2.3'.

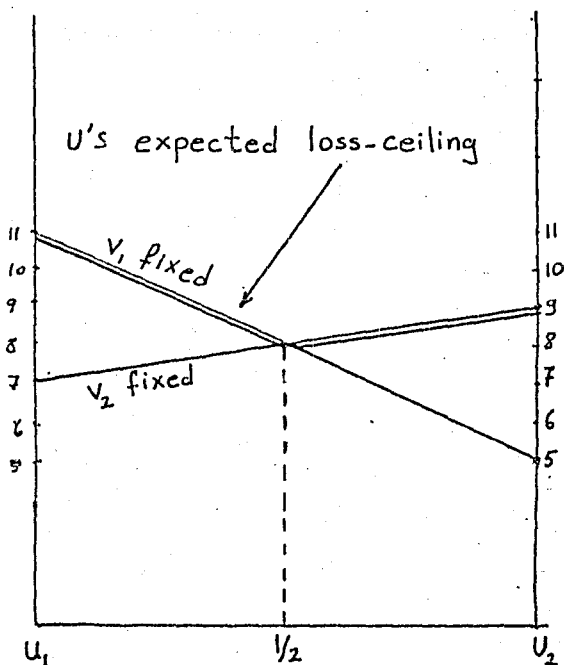


Figure 2.3. min-max-u plays first

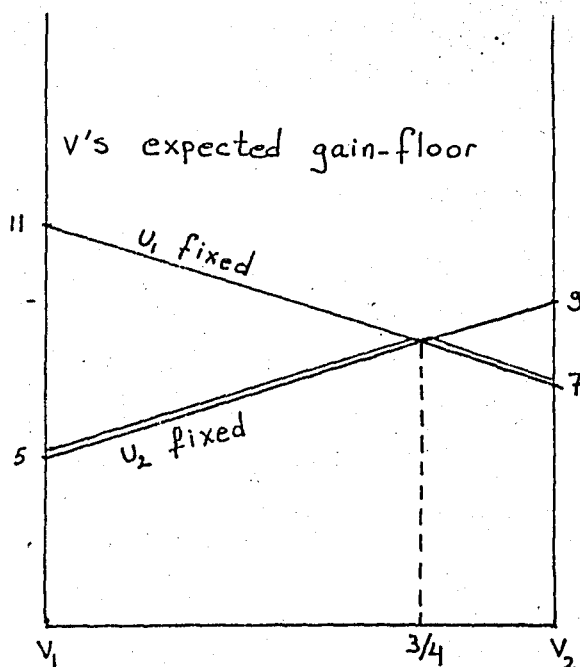


Figure 2.3'. max-min-v plays first

In figure 2.3. if U plays any probability mix other than half of time  $u_1$ , half of time  $u_2$ , V can obtain a higher average payoff by playing the fixed strategy indicated by the solid line. Similarly, we see that V must play the probability mix one-quarter of the time  $v_1$ , three quarters of the time  $v_2$  to realize the maximal expected payoff. It is no accident that

$$E \min_p \max_q L_{ij} = 8 = E \max_q \min_p L_{ij}$$

That is, by randomization the difference between minimax and maximin can be equalized on the expected value basis. In this case  $p^0 = 1/2$ ,  $q^0 = 3/4$ .

#### 2.4 COMPUTATION OF OPTIMAL STRATEGIES

Minimax theorem assures us that every two-person zero-sum game will have optimal strategies, but it does not give a hint as how to compute these optimal strategies.

##### Saddle points :

The simplest case occurs if a saddle point exists, i.e., if there exists an entry  $L_{ij}$  which is both the maximum entry in its column and the minimum in its row. In this case, the pure strategies  $i$  and  $j$ , or equivalently, the mixed strategies  $p$  and  $q$  with  $p_i = 1$ ,  $q_j = 1$  and all other components equal to zero, will be optimal strategies for players  $u$  and  $v$ , respectively.

##### Domination :

In a matrix  $L$ , we say the  $i^{\text{th}}$  row dominates the  $k^{\text{th}}$  row if

and  $L_{ij} \geq L_{kj}$  for all  $j$

$L_{ij} > L_{kj}$  for at least one  $j$

Similarly, we say the  $j^{\text{th}}$  column dominates the  $l^{\text{th}}$  column if

$L_{ij} \leq L_{il}$  for all  $i$

and

$L_{ij} < L_{il}$  for at least one  $i$ .

THEOREM :

Any optimal strategy for the game obtained by removing the dominated rows (or columns) will also be an optimal strategy for the original game.

2 x 2 Games: Suppose we are given the 2 x 2 matrix game

$$\underline{L} = \begin{bmatrix} l_{11} & l_{12} \\ l_{21} & l_{22} \end{bmatrix}$$

It may be that this game has a saddle point; if so, there is no problem. Suppose, however, that the game has no saddle point. It follows that the optimal strategies  $\underline{p} = (p_1 \ p_2)$  and  $\underline{q} = (q_1 \ q_2)$  must have positive components. Now, if the value of the game ( $V_I = V_{II} = V$ ) is  $V$  we have

$$l_{11}p_1q_1 + l_{12}p_1q_2 + l_{21}p_2q_1 + l_{22}p_2q_2 = V \quad 2.1$$

or

$$p_1(l_{11}q_1 + l_{12}q_2) + p_2(l_{21}q_1 + l_{22}q_2) = V \quad 2.2$$

The two terms in parenthesis are both less than or equal to  $V$ , since  $\underline{q}$  is by hypothesis an optimal strategy. Suppose one of them were less than  $V$ ; i.e., suppose

$$l_{11}q_1 + l_{12}q_2 < V \quad 2.3$$

$$l_{21}q_1 + l_{22}q_2 \leq V \quad 2.4$$

Then, since  $p_1 \geq 0$  and  $p_1 + p_2 = 1$  it follows that Eq. 2.2 would be strictly smaller than  $V$ . Then, it follows that the terms in parenthesis must be equal to  $V$ . Hence

$$\begin{aligned} l_{11}q_1 + l_{12}q_2 &= V \\ l_{21}q_1 + l_{22}q_2 &= V \end{aligned} \quad \text{or } \underline{L} \underline{q}^t = \begin{bmatrix} V \\ V \end{bmatrix} \quad 2.5$$

Similarly it can be seen that

$$\begin{aligned} l_{11}p_1 + l_{12}p_2 &= V \\ l_{21}p_1 + l_{22}p_2 &= V \end{aligned} \quad \text{or } \underline{p} \underline{L} = \begin{bmatrix} V & V \end{bmatrix} \quad 2.6$$

These equations together with the equations

$$p_1 + p_2 = 1 \quad \text{and} \quad q_1 + q_2 = 1 \quad 2.7$$

allows us to solve for  $\underline{p}$ ,  $\underline{q}$ , and  $V$ .

THEOREM :

Let  $\underline{L}$  be a  $2 \times 2$  matrix game. Then, if  $\underline{L}$  does not have a saddle point, its unique optimal strategies and value will be given by

$$\underline{q} = \frac{\underline{L}^* \underline{J}^t}{\underline{J} \underline{L}^* \underline{J}^t} \quad 2.8$$

$$\underline{p} = \frac{\underline{J} \underline{L}^*}{\underline{J} \underline{L}^* \underline{J}^t} \quad 2.9$$

$$V = \frac{|\underline{\underline{L}}|}{\underline{\underline{J}} \underline{\underline{L}}^* \underline{\underline{J}}^t} \quad 2.10$$

where  $\underline{\underline{L}}^*$  is the adjoint of  $\underline{\underline{L}}$ ,  $|\underline{\underline{L}}|$  is the determinant of  $\underline{\underline{L}}$ , and  $\underline{\underline{J}}$  is the vector  $[1 \ 1]$ .

## 2.5 CONTINUOUS GAMES

If the choices of  $U$  and  $V$  are continuous instead of discrete, there must be a continuous payoff function,  $L(u,v)$  instead of a payoff matrix  $L_{ij}$ . We look for a pair of choices,  $u^0, v^0$ , such that

$$L(u^0, v) \leq L(u^0, v^0) \leq L(u, v^0) \quad \text{for all } u, v$$

It is claimed that necessary conditions for  $u^0$  and  $v^0$  are

$$\frac{\partial L}{\partial u} = 0, \quad \frac{\partial L}{\partial v} = 0 \quad 2.11$$

$$\frac{\partial^2 L}{\partial u^2} \geq 0, \quad \frac{\partial^2 L}{\partial v^2} \leq 0 \quad 2.12$$

and sufficient conditions are Eq. 2.11 and Eq. 2.12 with the equalities changed to inequalities. Any  $u^0, v^0$  satisfying the sufficient conditions is called a "game-theoretic saddle point". It should be pointed out that the two equations are not equivalent to the usual conditions for a "calculus saddle point", which are

$$\frac{\partial L}{\partial u} = 0, \quad \frac{\partial L}{\partial v} = 0 \quad 2.11'$$

$$\frac{\partial^2 L}{\partial u^2} \frac{\partial^2 L}{\partial v^2} - \left( \frac{\partial^2 L}{\partial u \partial v} \right)^2 \leq 0 \quad 2.12'$$

## CHAPTER III

## GAME THEORETIC APPROACH

## 3.1 FORMULATION IN GAME SETTING

We can view the determination of a control law for an uncertain system as a zero-sum two-person game. The first player is the DESIGNER who must choose a feedback control law which generates the control function based on instantaneous observations of the states of the uncertain system. His opponent, referred to as NATURE, chooses the system parameters. When system performance is measured by a cost functional, the value of the cost functional is viewed as the designer's loss and nature's gain resulting from operation of the system viewed as the play of the game. The min-max criterion by which the control is chosen in a game formulation yields the smallest guaranteed upper bound on the cost. In the game formulation, the criterion used by the controller as a basis for determining the value of control to be applied not knowing  $\underline{v}$ , is

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v})$$

After this formulation in game setting the general min-max feedback control problem will be stated and formulated; then the formulas evolved will be applied to the specific case - linear, autonomous uncertain system.

## 3.2 STATEMENT OF THE MIN-MAX FEEDBACK CONTROL PROBLEM

In general, the uncertain dynamical system to be controlled may be modelled by the vector differential equation

$$\dot{\underline{x}}(t) = \underline{f}(\underline{x}(t), \underline{u}(t), \underline{v}; t) \quad , \quad \underline{v} \in V \quad 3.1$$

Here,  $\underline{x}(t)$  is an  $n$ -vector referred to as the state at time  $t$ ,  $\underline{u}(t)$  represents a vector of  $m$  forcing functions, and  $\underline{v}$  a set of  $p$  time invariant parameters.

The first participant in the game is the DESIGNER who is to choose a control law which generates  $\underline{u}(t)$  based on instantaneous observations of the state and time. Such a control is referred to as a memoryless or instantaneous feedback control and may be represented in the form

$$\underline{u}(t) = \underline{k}(\underline{x}(t), t) \quad 3.2$$

Denote by  $U$  the set of permissible feedback controls of the form Eq. 3.2 with elements  $\underline{u}$  mapping  $R^n \times [t_0, T]$  into  $R^m \times [t_0, T]$ .

The designer's opponent, referred to as NATURE, chooses the vector of parameters  $\underline{v}$  from the set  $V$ .

Associated with an initial condition

$$\underline{x}(t_0) = \underline{x}_0 \quad 3.3$$

for the system in Eq. 3.1 is a performance functional, denoted by  $J(\underline{u}, \underline{v}, \underline{x}_0, t_0)$  of the fixed time, free endpoint form

$$J(\underline{u}, \underline{v}; \underline{x}_0, t_0) = h(\underline{x}(T)) + \int_{t_0}^T l(\underline{x}(t), \underline{u}(t), t) dt \quad 3.4$$

$J(\underline{u}, \underline{v}, \underline{x}_0, t)$  is viewed as the controller's loss and nature's gain as a result of a play of the game corresponding to operation of the system. The criterion used by the controller, in the game theory formulation, was

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}, \underline{x}_0, t) \quad 3.5$$

Because a dynamic programming approach will be used, in order to avoid technical difficulties, it is convenient to assume that all functions involved in Eq. 3.1, Eq. 3.2 and Eq. 3.4 are continuous with respect to their arguments.

Note that the initial conditions  $\underline{x}_0$ , and to appear in Eq. 3.5 as specified constants. Hence, a solution to Eq. 3.5, denoted by the pair  $u^0, v^0$  will depend parametrically upon  $\underline{x}_0$ , and  $t_0$ . From a game viewpoint the initial conditions is treated as having been chosen by a neutral agent. Then Eq. 3.5 corresponds to the case where both the designer and nature base their decision upon knowledge of  $\underline{x}_0$ .

In seeking to apply techniques of deterministic optimal control to the min-max criterion used as a basis of choosing a control law for uncertain systems in the game setting, the major obstacle to be overcome is that the order of minimization and maximization cannot, in general, be interchanged, i.e., a min-max solution may exist but not a saddle point. In the theory of games (Ch. 2), the introduction of mixed strategies is used to create a saddle point solution.

### 3.3 INTRODUCTION OF MIXED STRATEGIES OVER THE UNCERTAINTY SET

Viewing the direct solution of the min-max problem (Eq. 3.5) as the minimization with respect to  $u$  of the functional  $\phi$  defined as

$$\phi(\underline{u}) = \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) \quad 3.6$$

indicates two sources of difficulty. First, evaluation of the functional  $\phi$  by maximization with respect to  $\underline{v}$  cannot usually be done in a convenient form. Secondly, if the maximizing  $\underline{v}$  is not unique, then  $\phi$  may not be differentiable and hence the minimum cannot be characterized as a stationary point of  $\phi$  with respect to  $\underline{u}$ .

Consider the interchange of the order of the minimization and maximization operations. The max-min problem

$$\max_{\underline{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \underline{v}) \quad 3.7$$

is readily solved by using deterministic optimal control to minimize with respect to  $u$  and, noting the parametric dependence of the optimal control on  $v$ , then maximizing over  $V$ .

From the game theory we know that

$$\max_{\underline{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \underline{v}) \leq \min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) \quad 3.8$$

For equality to hold in Eq. 3.8 a saddle point should exist. If it doesn't exist it is generated by introducing mixed strategies defined as the set of probability measures over the pure strategies, which are  $U$  and  $V$  in our case. Because of the measurability and implementation problems associated with introduction of randomized feedback, we will introduce mixed strategies only over the uncertainty set,  $V$ . Let  $\hat{V}$  denote the set of all probability measures over the compact set  $V$ . Elements in  $\hat{V}$  are denoted by  $\hat{v}$ . Note that the pure strategies are a subset of the mixed strategies corresponding to point distributions.

In the case of a finite uncertainty set

$$V = \{v_j : j = 1, \dots, N\} \quad 3.9$$

then the mixed strategies  $\hat{V}$  correspond to the simplex in  $R^N$  consisting of the set of all probability weights which satisfy the linear constraints.

$$\lambda_i \geq 0, \quad i = 1, \dots, N \quad \sum_{i=1}^N \lambda_i = 1 \quad 3.10$$

At this point one should state a very important lemma, called as equivalence lemma which states that min-max problem with mixed strategies over the uncertainty set is equivalent to the original min-max problem without mixed strategies.

Equivalence Lemma :

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) = \min_{\underline{u} \in U} \max_{\hat{v} \in V} J(\underline{u}, \hat{v})$$

Proof : Reference 1.

### 3.4 MIN-MAX THEOREM AND SOLUTION OF MODIFIED PROBLEM

The equivalence lemma assures that the min-max problem with mixed strategies over the uncertainty set is equivalent to the original min-max problem. Now a min-max theorem is needed which provides sufficient conditions such that solution of the modified problem

$$\max_{\hat{v} \in \hat{V}} \min_{\underline{u} \in U} J(\underline{u}, \hat{v})$$

MIN-MAX THEOREM Assume

- 1)  $J(\underline{u}, \underline{v})$  is continuous with respect to  $\underline{u}$  and  $\underline{v}$  for each  $\underline{u} \in U$  and  $\underline{v} \in V$ .
- 2)  $V$  is compact, i.e., closed and bounded.
- 3) For each  $\hat{v} \in \hat{V}$  there exists a unique minimizing  $\underline{u} \in U$  for the problem with random parameters

$$\min_{\underline{u} \in U} J(\underline{u}, \hat{v})$$

3.11

denoted by  $\underline{u}^0(\hat{v})$  which depends continuously on  $\hat{v}$ .

Then there exists a saddle point  $\underline{u}^* \in U$  and  $\underline{v}^* \in \hat{V}$  for  $D(\underline{u}, \underline{v})$  over  $U \times \hat{V}$  and

$$\begin{aligned} \min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) &= \max_{\underline{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \underline{v}) & 3.12 \\ &= J(\underline{u}^*, \underline{v}^*) \end{aligned}$$

Proof : Appendix C.

We can now examine the properties of the modified problem

$$\max_{\underline{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \underline{v}) \quad 3.13$$

It is clear that min-max theorem gives sufficient conditions for both existence of a solution to this problem and its equivalence to the min-max problem defined at Eq. 3.5. The dynamic programming viewpoint is applied to determine a necessary condition for a feedback control to solve the problem

$$\min_{\underline{u} \in U} J(\underline{u}, \underline{v}) \quad 3.14$$

viewed as an optimal control problem of minimizing the expected cost for a system with constant random parameters described by the arbitrary but fixed probability distribution  $\underline{v}$ . As a result an integro-differential equation which is analogous to the Hamilton-Jacobi equation for the corresponding deterministic control to be a solution of Eq. 3.14. in terms of this equation.

To obtain the necessary condition, we assume existence of a minimizing feedback control  $\underline{u}^0 \in U$  of the form

$$\underline{u}^0 = \underline{k}(\underline{x}(t), t) \quad 3.15$$

for the random parameter expected cost problem for all  $t_0 \leq t \leq T$  and all  $\underline{x}_0$ . Next, we define a real valued function  $S(\underline{x}, t; \underline{v})$  for all  $t_0 \leq t \leq T$ , all  $\underline{x}$ , and all  $\underline{v} \in V$  by

$$S(\underline{x}, t; \underline{v}) = J(\underline{u}^0, \underline{v}; \underline{x}, t) \quad 3.16$$

The interpretation of  $S(\underline{x}, t, \underline{v})$  is that it denotes the value of the cost functional resulting from using the feedback control which is optimal for the expected cost,  $\underline{u}^0$ , with a system whose parameters are actually equal to  $\underline{v}$  over the interval  $[t, T]$  and starting at point  $\underline{x}$ . Note that  $S(\underline{x}, t, \underline{v})$  does not therefore denote an optimal value function for any system. The optimal value function for the expected cost functional is however given in terms of  $S$  by expectation of  $S$ , i.e.,

$$\int_V S(\underline{x}_0, t_0; \underline{v}) d\underline{v} = \min_{\underline{u} \in U} J(\underline{u}, \hat{\underline{v}}; \underline{x}_0, t_0) \quad 3.17$$

This may be verified from Eq. 3.4. and assumption that  $\underline{u}^0$  is optimal.

For arbitrary points  $t_1$  and  $t_2$  satisfying  $t_0 \leq t_1 \leq t_2 \leq T$  the dynamic programming argument applied to the minimum cost function for the expected cost problem (Eq. 3.16) implies

$$\int_V S(\underline{x}_1, t_1; \underline{v}) d\underline{v} = \min_{\underline{k}(\underline{x}(t), t)} \int_V \left\{ \int_{t_1}^{t_2} l(\underline{x}, \underline{k}) d\tau + S(\underline{x}(t_2), t_2; \underline{v}) \right\} d\underline{v} \quad 3.18$$

where  $\underline{k}(\underline{x}(t), t)$  is written in place  $\underline{u}$  to emphasize a feedback control, as indicated in Eq. 3.15, is being sought.  $\underline{x}(t_2)$  is the result of the transition of the system from the state  $\underline{x}_1$  at  $t_1$  via the equation.

$$\dot{\underline{x}} = f(\underline{x}, \underline{k}, \underline{v}, t) \quad 3.19$$

Integrating along the trajectory determined by Eq. 3.19 we may write

$$S(\underline{x}(t_2), t_2; \underline{v}) = S(\underline{x}_1, t_1, \underline{v}) + \int_{t_1}^{t_2} \left\{ \frac{\partial S}{\partial \underline{x}} \underline{f} + \frac{\partial S}{\partial t} \right\} dz \quad 3.20$$

Substituting Eq. 3.20 into Eq. 3.19 we get

$$\int_V S(\underline{x}_1, t_1, \underline{v}) d\underline{v} = \min_{\underline{k}(\underline{x}(t), t)} \int_V \left\{ S(\underline{x}_1, t_1; \underline{v}) + \int_{t_1}^{t_2} \left( 1 + \frac{\partial S}{\partial \underline{x}} \underline{f} + \frac{\partial S}{\partial t} \right) dz \right\} d\underline{v} \quad 3.21$$

From Eq. 3.17 we conclude that the term on the right side of Eq. 3.21 given by

$$S(\underline{x}_1, t_1, \underline{v}) d\underline{v}$$

is independent of  $\underline{k}$  and hence may be cancelled from both sides of Eq. 3.20 to leave

$$0 = \min_{\underline{k}} \int_V \int_{t_1}^{t_2} \left\{ 1 + \frac{\partial S}{\partial \underline{x}} \underline{f} + \frac{\partial S}{\partial t} \right\} dz d\underline{v} \quad 3.22$$

Under general conditions of integrability the order of integration in Eq. 3.22 may be reversed by Fubini's theorem to obtain

$$0 = \min_{\underline{k}} \int_{t_1}^{t_2} \int_V \left\{ 1 + \frac{\partial S}{\partial \underline{x}} \underline{f} + \frac{\partial S}{\partial t} \right\} d\underline{v} dz \quad 3.23$$

$1, \underline{f}$ , and  $\underline{k}$  with respect to all their arguments are continuous which implies the continuity of the integrand with respect to  $z$  along trajectories associated with the optimal feedback for all  $\underline{v}$ . This implies continuity with respect to  $z$  of the inner integral and hence applying mean value theorem to eliminate the integral over  $[t_1, t_2]$  we obtain that for some  $t \in [t_1, t_2]$

$$0 = \min_{\underline{k}} \int_V \left\{ 1 + \frac{\partial S}{\partial \underline{x}} \underline{f} + \frac{\partial S}{\partial t} \right\} d\underline{v} \quad 3.24$$

Since  $t_1$  and  $t_2$  were arbitrary, we conclude that Eq. 3.24 is a necessary condition which the optimal control  $\underline{u}^0$  must satisfy for all  $t$  in  $[t_0, T]$

If the expression

$$\int_V \left\{ l(\underline{x}, \underline{k}, t) + \frac{\partial S}{\partial \underline{x}}(\underline{x}, t, \underline{v}) f(\underline{x}, \underline{k}, \underline{v}, t) \right\} d\underline{v} \quad 3.25$$

which is the expected Hamiltonian has the property of having a unique minimum with respect to  $\underline{k}$  for all  $\underline{x}$  and all  $t \in [t_0, T]$ , then the problem is said to be normal in analogy with the deterministic case. Let  $\underline{k}^0$  denote the unique minimizer in the normal case. Then by substituting  $\underline{k}^0$  into Eq. 3.24 we obtain the analogous of Hamilton-Jacobi equation.

$$\int_V \left\{ l(\underline{x}, \underline{k}^0(\underline{x}, t)) + \frac{\partial S}{\partial \underline{x}}(\underline{x}, t, \underline{v}) f(\underline{x}, \underline{k}^0(\underline{x}, t, \underline{v})) + \frac{\partial S}{\partial t}(\underline{x}, t, \underline{v}) \right\} d\underline{v} = 0 \quad 3.26$$

which is an integro-differential equation which must be satisfied by  $S(\underline{x}, t, \underline{v})$

The sufficient condition for a control  $\underline{k}^0(\underline{x}, t)$  to be optimal can now be stated in terms of Eq. 3.26 in analogy with the Hamilton-Jacobi sufficiency condition for deterministic optimal control problem as follows. If Eq. 3.25 is normal and a unique solution  $S(\underline{x}, t, \underline{v})$  can be found satisfying the integro-differential equation (Eq. 3.26), then a solution to problem

$$\min_{\underline{u} \in U} J(\underline{u}, \underline{v})$$

is found which is unique except on a set of measure zero.

In the following section these results are applied to the linear system quadratic criterion case where  $S(\underline{x}, t, \underline{v})$  separates into the quadratic form  $\underline{x}^T \underline{R}(t, \underline{v}) \underline{x}$

## 3.5 THE LINEAR SYSTEM, QUADRATIC CASE

We begin by specializing the results of the previous section to the case of a linear system, described by setting

$$f(\underline{x}, \underline{u}, \underline{v}) = \underline{A}(\underline{v})\underline{x} + \underline{B}(\underline{v})\underline{u} \quad 3.27$$

and quadratic criterion given by

$$l(\underline{x}, \underline{u}) = 1/2 \underline{x}^T \underline{Q} \underline{x} + 1/2 \underline{u}^T \underline{P} \underline{u} \quad 3.28$$

where  $\underline{Q}$  is nonnegative definite and  $\underline{P}$  is positive definite.

As in the deterministic case assume a quadratic form for  $S$  given by

$$S(\underline{x}, \underline{v}) = 1/2 \underline{x}^T \underline{R}(\underline{v})\underline{x} \quad 3.29$$

Note that the assumptions on  $\underline{Q}$  and  $\underline{P}$  imply that  $J(\underline{u}, \underline{v})$  is nonnegative for all  $\underline{v} \in V$ ,  $\underline{u} \in U$ , and hence the matrix  $\underline{R}(\underline{v})$  may be assumed symmetric and nonnegative definite.

On analogy to the deterministic case, the following Riccati integro-differential equation for  $\underline{R}(\underline{v})$  is obtained, where the bar notation denotes expectation with respect to  $\mathcal{V}$  over  $V$ .

$$\begin{aligned} \dot{\underline{R}}(\underline{v}) = & \underline{A}^T \underline{R} - \underline{R} \underline{A} + (\underline{B}^T \underline{R})^T \underline{P}^{-1} \overline{\underline{B}^T \underline{R}} + \overline{\underline{B}^T \underline{R}} \underline{P}^{-1} \underline{B}^T \underline{R} \\ & - \underline{Q} - (\underline{B}^T \underline{R})^T \underline{P}^{-1} \overline{\underline{B}^T \underline{R}} \quad \underline{v} \in V \end{aligned} \quad 3.30$$

Minimization of the Hamiltonian yields a unique solution

$$\underline{u}^0 = - \underline{P}^{-1} \overline{\underline{B}^T \underline{R}} \underline{x}(t) \quad 3.31$$

By normality of the problem, the existence and uniqueness of the minimizing feedback control is equivalent to existence and uniqueness of a solution to Riccati integro-

differential equation.

In case of a finite uncertainty, set  $\hat{v}$  is represented by probability weights satisfying

$$\lambda_i > 0, \sum_{i=1}^N \lambda_i = 1$$

Then Riccati - Integro - differential equation (Ex.3.30) reduces to a set of N-coupled Riccati Differential equations with parameters .

Now two theorems about existence and uniqueness of solution will be stated without proof. For proofs one may refer to reference 1.

THEOREM :

The minimum expected cost feedback control for the case of a linear system, quadratic criterion finite uncertainty set problem exists, is unique and depends continuously on  $\hat{v}$ . It is given by

$$\underline{u}^0 = - \underline{P}^{-1} \underline{B}^T \underline{R} \underline{x}(t)$$

in terms of a solution of the coupled Riccati-differential equations (Eq. 3.30).

THEOREM :

In the case of a finite uncertainty set, a solution exists to the linear system quadratic criterion min-max control problem. The feedback law and min-max value are given by Eq. 3.31 and Eq. 3.29, respectively, and in terms of the solution of the coupled Riccati differential equations (3.30). The min-max feedback law is linear except for parametric dependence upon initial state  $\underline{x}_0$ .

In the problem considered in this study

$$t_0 = 0 \quad \text{and} \quad T \rightarrow \infty$$

and that's why, differential term  $\dot{R}(\underline{y})$  drops and Eq. 3.30 becomes simpler, called Algebraic Riccati Equation (ARE).

### 3.6 A SADDLE POINT TEST

In many min-max problems the hypotheses of a min-max theorem providing sufficient conditions for existence of a saddle point cannot be verified. Especially in such situations it should be pointed out that if a max-min solution is obtainable, then it may be possible to apply the following test giving necessary and sufficient conditions for the solution to be a saddle point and hence also a min-max solution.

#### Saddle Point Test:

Let  $\hat{V}_1$  denote an arbitrary subset of  $\hat{V}$  and assume  $\underline{u}' \in U$  and  $\underline{\hat{v}}' \in \hat{V}_1$  are a solution to the max-min problem

$$\max_{\underline{\hat{v}} \in \hat{V}_1} \min_{\underline{u} \in U} J(\underline{u}, \underline{\hat{v}}) \quad 3.32$$

Then  $\underline{u}', \underline{\hat{v}}'$  is a saddle point for  $J(\underline{u}, \underline{\hat{v}})$  over  $U \times \hat{V}$  and hence also a solution to min-max problem

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) \quad 3.33$$

if and only if

$$\max_{\underline{v} \in V} J(\underline{u}', \underline{v}) = J(\underline{u}', \underline{\hat{v}}') \quad 3.34$$

PROOF : Reference 1

### 3.7 PROCEDURE FOR SOLVING THE MIN-MAX PROBLEM IN GAME THEORETIC APPROACH

In this thesis study the following procedure has been used to solve the min-max feedback control problem in a game theory setting.

- 1) Solve the max-min problem without introduction of mixed-strategies.

To solve the max-min problem, one first finds the solution to the deterministic optimal control problem with parameter  $\underline{v}$  obtained by solution of the Riccati equation. Then one maximizes the quadratic cost functional  $J$  which is given by

$$J = \underline{x}_0^T \underline{R}(\underline{v}) \underline{x}_0 \quad 3.35$$

over the set of parameter  $\underline{v}$ ; and note the max-min value.

- 2) Apply Saddle-Point Test

- a) Stability test

First, to test the condition of the saddle point test, it is noted that for an arbitrary constant feedback matrix  $\underline{K}$  such that closed loop system

$$\dot{\underline{x}} = (\underline{A} + \underline{B} \underline{K}) \underline{x} \quad 3.36$$

is stable. In this case, the quadratic cost is given by Ex. 3.35, but now,  $R$  is the symmetric nonnegative definite solution of the equations (Liapunov Equations) below

$$(\underline{A} + \underline{B} \underline{K})^T \underline{R} + \underline{R}(\underline{A} + \underline{B} \underline{K}) = \underline{Q} - \underline{K}^T \underline{P} \underline{K} \quad 3.37$$

- b) In place of  $\underline{K}$  matrix in above equation the max-min

control vector (or matrix) is replaced, and Liapunov equation is solved. The solution of Liapunov equations depend parametrically on  $v$ ; and maximization of the cost functional  $J$  over the set of parameter  $v$  will give us a new value. This new value is compared with the max-min value found at step 1 of the procedure.

3) If the max-min value of step 1 and the new value found at step 2 are equal, it means that there exists a saddle point and max-min value equals min-max value.

If, however, those two values are not equal, it means that mixed strategies should be introduced or as it is done in this study successive search method can be used.

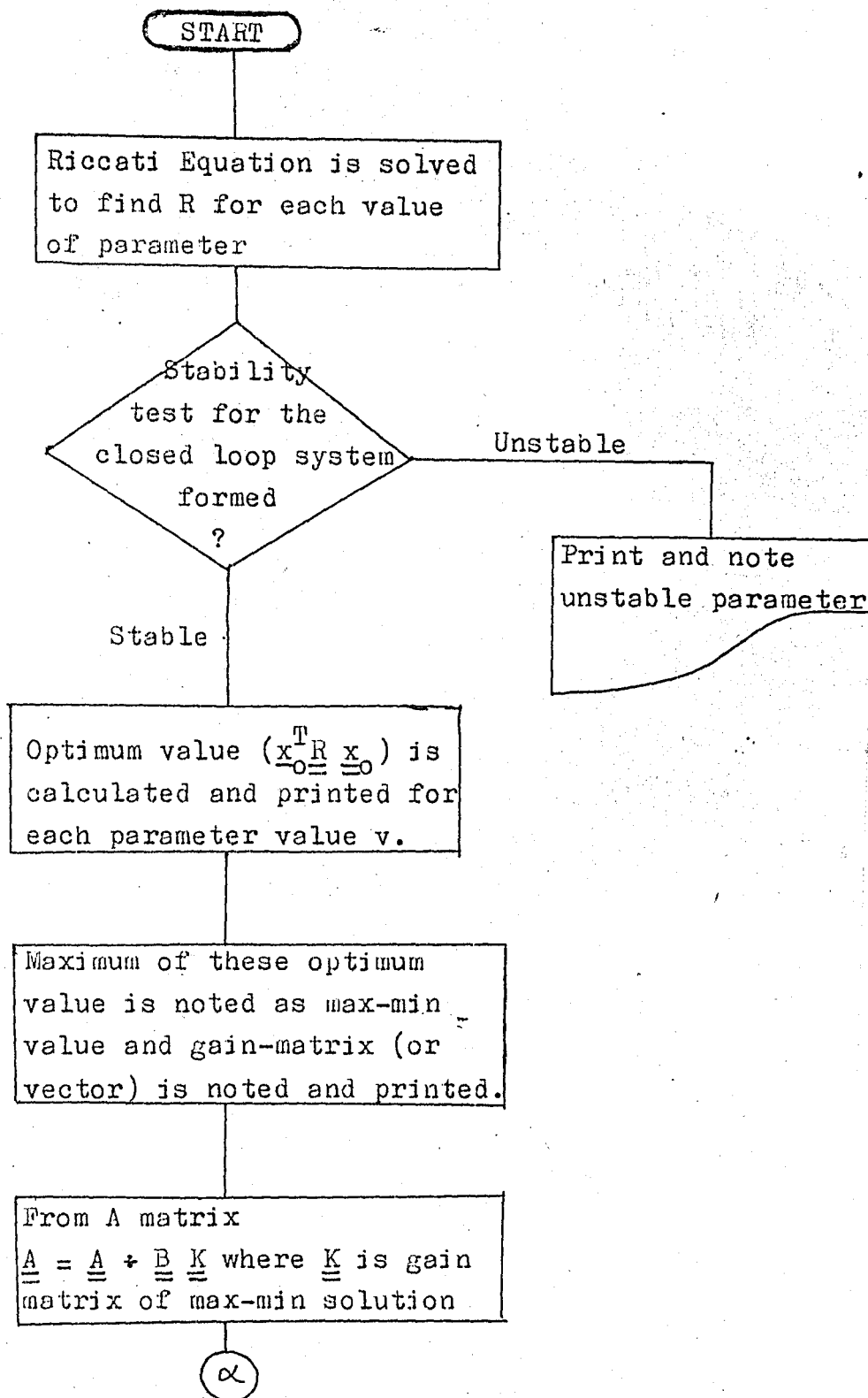
As it is clear from the above procedure the two matrix equations, namely, Algebraic Riccati and Liapunov equations, have to be solved in computer in order to find the max-min solution and make the saddle-point test. Another point which should be pointed out in the above procedure is that the cost functional  $J$  in step 2b is again given by

$$J = \underline{x}_0^T \underline{R} \underline{x}$$

as in step 1; however,  $R$ , in this case, is the Liapunov equation solution. This point is a claim which needs at least a justification. This justification will be given in the appendix part of this thesis.

Now, a flow-chart of this game theoretic approach method will be given in order to give an overview of the computer program.

## 3.8 FLOW - CHART OF THE METHOD



(α)

Solve the Liapunov eqn.  

$$(\underline{A} + \underline{B} \underline{K})^T \underline{R} + \underline{R}(\underline{A} + \underline{B} \underline{K}) - \underline{Q} - \underline{K} \underline{P} \underline{K}$$

Calculate cost functional  
 $(\underline{x}_0^T \underline{R} \underline{x}_0)$  and note the maximum of these values over the parameter set V.

Compare  
 the last maximum  
 value with the max-min  
 value. MAX-MIN  
 ZOROLD

EQUAL

SADDLE-POINT EXISTS  
 Min-Max Value equal  
 Max-Min Value

NOT EQUAL

STOP

SADDLE POINT DOES NOT  
 EXIST: Search Method  
 Should be Used.

### 3.9 DISCUSSION

The most important aspect of this method is that if there exists a saddle point the solution is found out rather rapidly. For instance, the program runs for 1 minute 38 seconds for one parameter uncertainty (11 grid points), and it runs for 1 minute 42 seconds for two parameter (16 grid points), and it runs for 3 minutes 7 seconds for three parameters (64 grid points). If, however, the saddle point does not exist, then it is faced with a rather difficult problem of introduction of mixed strategies. In this study, this problem has been skipped and search method has been used since there had been a difficulty in choosing the parameter set to introduce mixed strategies. Thus, mixed strategies would be a new area for those who are interested; and hopefully this thesis will be a good reference and hopefully a basis for those people.

## CHAPTER IV

## GRID POINT SEARCH METHOD

## 4.1 PHILOSOPHY OF THE METHOD

One classical approach to the problem of uncertainty in the design of control systems is to introduce a sufficient amount of feedback so as to make the system insensitive to variations in plant parameters or controllers. However, the mere incorporation of negative feedback around the plant does not guarantee for which the controlled quantity is unaffected by the uncertainty.

Since our aim is to design an insensitive system against various variations in plant we have to have a definition of sensitivity. The sensitivity of a control system is usually taken to be the normalized variation of some desired characteristic with the variation of plant or controller parameters. Let us assume that  $T$  is desired quantity and  $v$  is the parameter; then the sensitive is defined to be

$$S_v^T = \frac{\Delta T}{\Delta v} \frac{v}{T} \quad 4.1$$

This form of sensitivity gives a measure of the normalized deviation from desired value  $T$ . The above description is an absolute sensitivity definition. In this search method the absolute sensitivity measure has been used. As it is discussed in chapter 1 section 2 it has been searched for  $u^0$  and a parameter  $v^0$  such that

$$|J(u^0, v^0) - J(u, v)| \leq \max |J(u, v^0) - J(u, v)| \text{ for all } u \in F^0$$

4.2

Since the whole idea behind this method is presented in section 2 of Chapter 1 now the procedure will be considered.

#### 4.2 PROCEDURE FOR GRID-POINT SEARCH

Let us assume that there are two parameters  $v_1, v_2$  subject to the variations

$$0.8 \leq v_1, v_2 \leq 1.2$$

The two-dimensional parameter space is divided into twenty-five grid points. The spacing is 0.1 on each axis, that is, we will allow for 0.1 changes in the two parameters  $v_1, v_2$ .

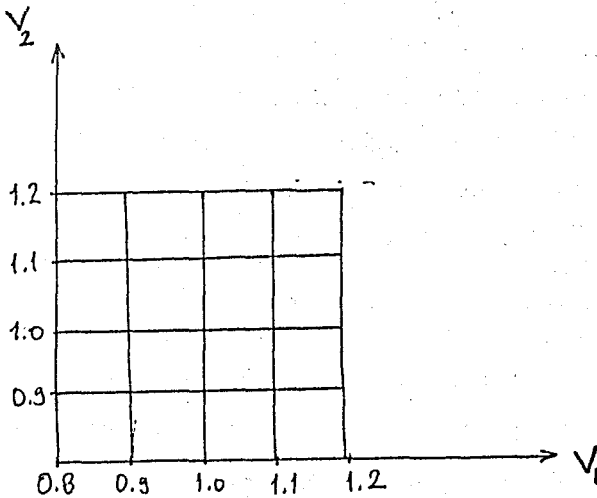


Figure 4.2. Parameter grid

The procedure consists of the following steps.

- 1) The optimal control law for a particular grid point is evaluated using the principles of optimal control theory. For optimality,  $u$  is given as a function of the state variables

$$\underline{u} = - \underline{P}^{-1} \underline{B}^T \underline{R} \underline{x}$$

where  $\underline{R}$  satisfies the matrix equation (ARE).

$$\underline{A}^T \underline{R} + \underline{R} \underline{A} - \underline{R} \underline{B} \underline{P}^{-1} \underline{B}^T \underline{R} + \underline{Q} = 0$$

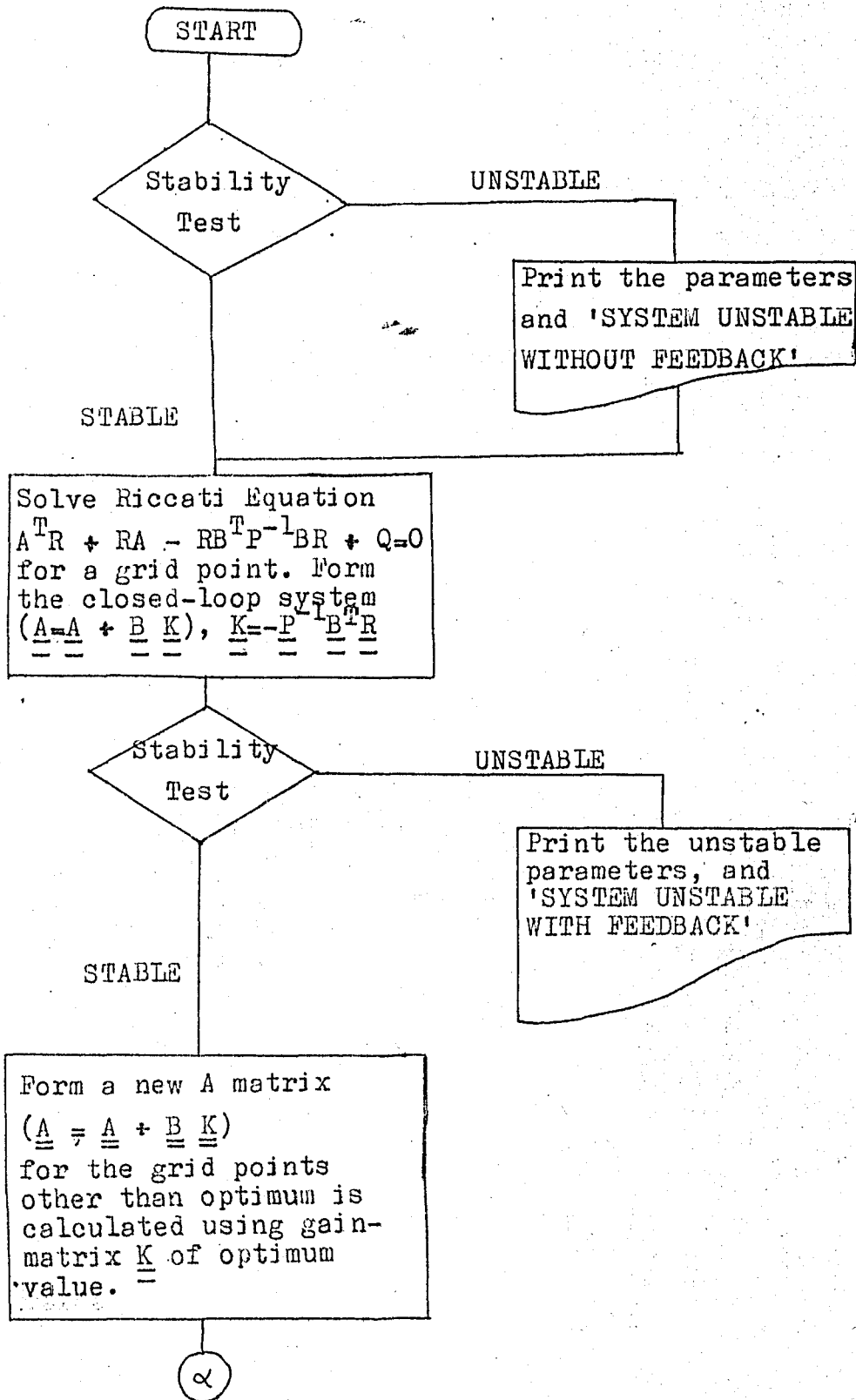
(The equation is algebraic, since the system is autonomous.)

2) Taking the optimal control law for the grid point (1,1), the performance indices at all other points are evaluated (this corresponds  $J_2^1$  in Fig. 1.2.1) and deviations from their respective optima are calculated (corresponding to the difference ( $J_2^1 - J_2^0$ )). The maximum of these deviations is noted. Now taking the optimal control for the grid point (1,2), the calculations are repeated and the maximum deviation is found again. The procedure is repeated for all the other grid points. The control vector that yields the minimum of these maximum deviations is the one that will be chosen for the system, since it ensures the least drastic effects on the performance.

In the study of Sudhakar and Eyman (Ref 2) the procedure was developed for two parameter uncertainty. Here in this study a program has been developed up to five parameter uncertainty case.

The flowchart of the computer program of this method is given on the next page.

## 4.3 FLOW-CHART OF GRID-POINT SEARCH



(α)

Solve the Liapunov Eq.  

$$\underline{A}^T \underline{R}_L + \underline{R}_L \underline{A} = -\underline{Q} - \underline{K}^T \underline{P}^{-1} \underline{K}$$
 to find  $\underline{R}_L$ .

Calculate the performance value using Liapunov eq. solution  $\underline{x}_0^T \underline{R}_L \underline{x}_0$

Find the difference between optimum value and the performance value calculated above.

Note and print the maximum of these deviations

Note and print the minimum of those maximum values

#### 4.4 DISCUSSION AND COMPARISON

This grid-point search method is used when the saddle point solution does not exist; since it directly gives the min-max solution. This is the main advantage of this method.

However, one great disadvantage of this method when compared to game theoretic approach is that run-time (CPU) of the computer program of this search method is almost twice of the game theoretic approach method.

For instance, it takes 7 minutes 33 seconds for three parameter uncertainty problem (64 grid points); whereas the same problem is solved in 3 minutes 7 seconds by the game theoretic approach method.

## CHAPTER V

EXAMPLE PROBLEM SOLUTIONS  
IN TWO METHODS

## 5.1 ONE PARAMETER UNCERTAINTY CASE

Consider the determination of the min-max feedback control for the uncertain second order stationary linear system described by

$$\ddot{x} + 2w\zeta\dot{x} + w^2x = u$$

We know from control theory if  $0 < \zeta < 1$ , the closed-loop poles are complex conjugates and lie in the left-half S-plane. The system is then called underdamped, and transient response is oscillatory.

Suppose that  $\zeta = 0.5$ , and the variation in the natural frequency term  $w$  is as follows

$$0.2 \leq w \leq 2.2$$

with initial conditions

$$x_0 = 1, \dot{x}_0 = 0$$

and the quadratic cost functional

$$J(u, v) = \int_0^{\infty} (\dot{x}^2 + x^2 + u^2) dt$$

In the state-variable formulation corresponding to letting

$$\underline{x} = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} x \\ \dot{x} \end{bmatrix}$$

we have

$$\underline{A} = \begin{bmatrix} 0 & 1 \\ -v_1^2 & -v_1 \end{bmatrix} \quad \underline{b} = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad Q = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \quad P=1 \quad \underline{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

that is,

$$\dot{\underline{x}} = \begin{bmatrix} 0 & 1 \\ -v_1^2 & v_1 \end{bmatrix} \underline{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \quad \underline{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$P.I = \int_0^{\infty} (\underline{x}^T \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \underline{x} + u^2) dt$$

#### A) SOLUTION OF GAME THEORETIC APPROACH METHOD

MAX-MIN VALUE = 31.6

PARAMETER VALUE = 0.2

SADDLE POINT DOES NOT EXIST

#### B) SOLUTION OF GRID POINT SEARCH

PARAMETER VALUE = 1.6

### 5.2 TWO PARAMETER UNCERTAINTY CASE

In this case I have considered a second order plant with two parameter uncertainty described by

Plant:

$$\dot{\underline{x}} = \begin{bmatrix} 0 & 1 \\ -v_1 & -v_2 \end{bmatrix} \underline{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u, \quad \underline{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

Performance Index :

$$J(u, v) = \int_0^{\infty} (\underline{x}^T \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \underline{x} + u^2) dt$$

Variation in parameters

$$1.0 \leq v_1, v_2 \leq 2.0$$

grid is set up as in figure 5.2. below

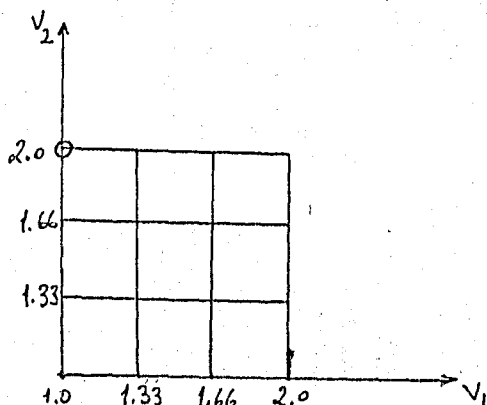


Figure 5.2. Parameter grid.

#### A) SOLUTION OF GAME THEORETIC APPROACH METHOD

$$\text{MAX-MIN VALUE} = 1.41$$

PARAMETER VALUES :

$$v_1 = 1.0, v_2 = 2.0$$

$$\text{GAIN VECTOR } \underline{K} = \begin{bmatrix} 0.41 \\ -0.41 \end{bmatrix}$$

SADDLE POINT DOES NOT EXIST

#### B) SOLUTION OF GRID-POINT SEARCH METHOD

PARAMETER VALUES

$$v_1 = 1.0, v_2 = 2.0$$

Eventhough, those values for the parameters coincide in min-max and max-min solution it does not require a saddle-point.

### 5.3 THREE PARAMETER UNCERTAINTY CASE

In this case I have considered again a second order system with three parameter uncertainty which is described by

plant:

$$\dot{\underline{x}} = \begin{bmatrix} 0 & v_3 \\ -v_1 & -v_2 \end{bmatrix} \underline{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u \quad \underline{x}_0 = \underline{x}(0) = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

Performance Index given by

$$J(u, v) = \int_0^{\infty} (\underline{x}^T \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \underline{x} + u^2) dt$$

and parameters variation

$$1.0 \leq v_1, v_2, v_3 \leq 2.0$$

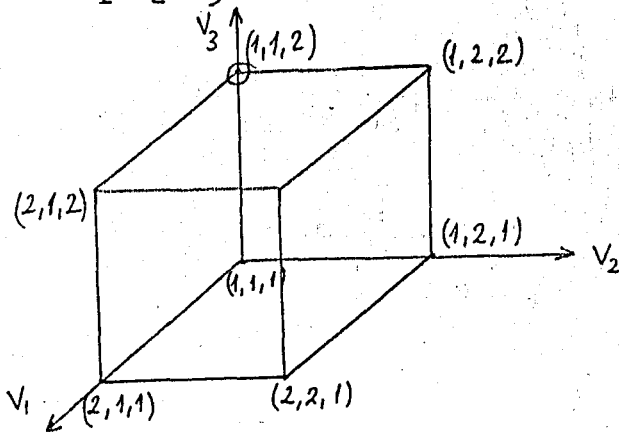


Fig. 5.3 Grid-points for three parameters

Grid is set up as in figure 5.3 above.

#### A) SOLUTION OF GAME THEORETIC APPROACH METHOD

MAX-MIN VALUE = 1.51

PARAMETER VALUES

$$v_1 = 1.0, v_2 = 1.0, v_3 = 2.0$$

$$\text{GAIN VECTOR } \underline{K} = \begin{bmatrix} 0.23 \\ -0.57 \end{bmatrix}$$

## b) SOLUTION OF GRID-POINT SEARCH METHOD

PARAMETER VALUES

$$v_1 = 1.0 , v_2 = 1.0 , v_3 = 2.0$$

## CHAPTER VI

### CONCLUSION AND COMPARISON OF TWO METHODS

This study has revealed that when the saddle point does not exist it is better to apply a search method over the uncertain parameter set; since search methods directly gives the min-max solution. When the saddle point exists, it is wiser to apply the game theoretic approach method. However, if one shows that even after the introduction of the mixed strategies over the uncertain parameter set computer program of the game theoretic approach method takes shorter time than the search method, then it may be concluded that game theoretic approach is preferable for the two cases - saddle - point exist or saddle - point does not exist.

If the system under consideration is a non-linear one, then it is again preferable to use game theoretic approach method since integro-differential equation (Eq. 3.24) which is analogous to Hamilton-Jacobi Equation of Control theory is ready for use. To apply the search method in the case of a nonlinear plant requires further development.

Thus, introducing mixed strategies over the uncertain parameter set in case of nonexistence of saddle-point will be a further study area. The contribution of such a study will be great since it will give an answer to question of computer time after the introduction of mixed strategies.

For both grid-point search or game theoretic approach it is true that as one wants to be more accurate one needs more grids points which in return implies more computer time. In this study no criteria has been developed in choosing the number of grid points. This may also be a further study area together with introduction of mixed strategies over the uncertain parameter set.

It is also true as the number of uncertain parameters increases it becomes harder and harder to be accurate because run-time (CPU) in computer increases very rapidly. In going from 2 parameter uncertainty case to 3 parameter uncertainty case with the same accuracy one needs four times as much computer time, because the number of grid-points increases geometrically.

## APPENDIX A

NUMERICAL SOLUTION OF ALGEBRAIC  
MATRIX RICCATI EQUATION

## A.1 METHODS OF SOLVING (ARE).

The quadratic matrix equation

$$\underline{\underline{A}} \underline{\underline{X}} + \underline{\underline{X}} \underline{\underline{A}}^T - \underline{\underline{X}} \underline{\underline{B}} \underline{\underline{X}} + \underline{\underline{C}} = \underline{\underline{0}} \quad \text{Eq. A1}$$

arises in many diverse application - steady state optimal control and filtering, radiative transfer and neutron diffusion, and economic equilibria, to name a few. Eq. A1, widely known as the Algebraic Riccati Equation (ARE), is deceptively simple in appearance; however, considerable effort has gone into development of effective numerical procedures for solving the Riccati equation A1. Upto now six solution methods actually have been used.

(i) Integrate the Riccati differential equation

$$\underline{\underline{\dot{X}}} = \underline{\underline{A}} \underline{\underline{X}} + \underline{\underline{X}} \underline{\underline{A}}^T - \underline{\underline{X}} \underline{\underline{B}} \underline{\underline{X}} + \underline{\underline{C}} \quad \text{to steady state}$$

(ii) Integrate the "Hamiltonian" system of linear differential equations

$$\begin{bmatrix} \dot{X} \\ Y \end{bmatrix} = \underline{\underline{\mathcal{H}}} \begin{bmatrix} X \\ Y \end{bmatrix}$$

where

$$\underline{\underline{\mathcal{H}}} = \begin{bmatrix} \underline{\underline{A}}^T & -\underline{\underline{B}} \\ -\underline{\underline{C}} & -\underline{\underline{A}} \end{bmatrix}$$

to steady-state (Kalman and Englar (1966))

- (iii) Solve the linear system  $p(\mathcal{H}) \begin{bmatrix} I \\ X \end{bmatrix} = 0$  where  $\mathcal{H}$  is the "Hamiltonian" matrix defined above and  $p(s)$  is the polynomial whose zeros are the  $n$  eigenvalues of  $\mathcal{H}$  having negative real parts (Bucy and Joseph (1968)).
- (iv) Transform equation A1 to a (discrete Riccati) quadratic matrix equation and solve that (Hitz and Anderson (1972))
- (v) Calculate  $\underline{X} = \underline{U} \underline{V}^{-1}$  where  $\begin{bmatrix} \underline{U} \\ \underline{V} \end{bmatrix}$  consists of the eigenvectors of  $\underline{A}$ , corresponding to the  $n$  eigenvalues with negative real parts (Mc Farlane (1963) and Potter (1964)).
- (vi) Convert the quadratic matrix problem to a sequence of linear problems  $\underline{A} \underline{X} + \underline{X} \underline{A}^T + \underline{C} = 0$  where  $\underline{A} = \underline{A} - \underline{X} \underline{B}$  and  $\underline{C} = \underline{C} + \underline{X} \underline{B} \underline{X}$  and solve these iteratively (Kleinman (1968)).

Recent investigations suggest that method (v), eigenvalue-eigenvector method and method (vi), quasilinearization method, are effective general-purpose algorithms for solving the Riccati Matrix equation A1.

## A.2 SOME PROPERTIES OF SOLUTIONS TO (ARE).

In this section let us quote relevant results concerning properties to the Algebraic Riccati Equation (ARE). The basic existence theorem was established by Kalman. It is given below as theorem A1. First, we restate the Riccati equation A1 in a more suitable form. It is

$$\underline{A} \underline{X} + \underline{X} \underline{A}^T - \underline{X} \underline{B}^T \underline{B} \underline{X} + \underline{C} \underline{C}^T = 0$$

Eq. A2

where  $A, B$  and  $C$  are given matrices -  $(n \times n)$ ,  $(m \times n)$  and  $(n \times 1)$ , respectively, such that  $\underline{B}^T \underline{B}$  is positive definite and  $\underline{C} \underline{C}^T$  is nonnegative definite. Depending upon the conditions which  $A, B$  and  $C$  are required to satisfy there may be a unique solution, a finite number of solutions or infinitely many solutions.

Definition 1 An eigenvalue is Stable if its real part is negative and it is unstable otherwise.

Definition 2 The matrix pair  $(\underline{A}, \underline{B})$  is stabilizable if there exists an  $\underline{X}$  such that all eigenvalues of  $\underline{A} - \underline{X} \underline{B}^T \underline{B}$  are stable.

THEOREM A1 A necessary and sufficient condition for the Riccati matrix equation (A2) to possess a unique solution  $\underline{X}$ , for which all eigenvalues of  $\underline{A} - \underline{X} \underline{B}^T \underline{B}$  are stable, is that  $(\underline{A}, \underline{B})$  be stabilizable and  $(\underline{C}, \underline{A})$  be detectable.

Definition 3 An eigenvalue of  $\underline{A}$  is  $(\underline{A}, \underline{B})$  - controllable if  $\underline{X} \underline{B} = 0$  implies  $\underline{x} = 0$  for all (left) eigenvectors of  $\underline{A}$ ;  $\underline{x} \underline{A} = \underline{x}$ ; is  $(\underline{A}, \underline{B})$  - uncontrollable if  $\underline{x} \underline{B} = 0$  for a nontrivial eigenvector  $\underline{x}$  of  $\underline{A}$ .

Definition 4 The matrix pair  $(\underline{A}, \underline{B})$  is controllable if all unstable eigenvalues of  $\underline{A}$  are  $(\underline{A}, \underline{B})$  - controllable.

Definition 5 The matrix pair  $(\underline{C}, \underline{A})$  is detectable if  $(\underline{A}^T, \underline{C}^T)$  is controllable.

Detectability is thus the dual condition to controllability.

THEOREM A2 A sufficient condition for  $(\underline{A}, \underline{B})$  to be stabilizable is that  $(\underline{A}, \underline{B})$  be controllable.

In this thesis study I have used method (vi), quasilinearization method of Kleinman (1968).

### A.3 QUASILINEARIZATION METHOD

This method is essentially a Newton-raphson procedure. It proceeds as follows

(i) Choose an initial  $\underline{X}$  such that  $\underline{A} = \underline{A} - \underline{X} \underline{B}$

(ii) Solve the Liapunov matrix equation

$$\underline{A} \underline{X} + \underline{X} \underline{A}^T + \underline{C} = 0$$

where  $\underline{C} = \underline{C} + \underline{X} \underline{B} \underline{X}$  using the previous  $\underline{X}$ .

(iii) If the relative change in  $\underline{X}$  is sufficiently small and if the error  $\underline{A} \underline{X} + \underline{X} \underline{A}^T - \underline{X} \underline{B} \underline{X} + \underline{C}$  is sufficiently small, then exist. Otherwise, return to step (ii).

Notice that, under the assumptions necessary to assure the existence of a unique solution to the Riccati matrix equation, an initial stabilizing  $\underline{X}$  exists. Moreover, the sequence of solutions to the Liapunov equation in step (ii) converges monotonically to the solution of the Riccati Equation (A1).

To become a competitive numerical method for solving the (ARE), the quasilinearization algorithm requires efficient numerical procedures for stabilization, that is, computing an initial  $\underline{X}$  in step (i), and for solving the Liapunov matrix equation in step (ii).

The computer program for solving Riccati Matrix equations employs a stabilization procedure as follows. If (A1) is controllable, then an initial stabilizing  $\underline{X}$  is given by  $\underline{X} = \underline{W}^{-1} \underline{B}$  where  $\underline{W}$  is the "observability" matrix

$$\underline{W} = \int_0^T e^{-\underline{A}s} \underline{B} e^{-\underline{A}^T s} ds$$

for some  $T > 0$ . A Simpson quadrature algorithm is used to calculate  $\underline{W}$ . This involves choosing a stepsize  $h$  and approximating the integral by

$$\underline{W} = \left( \frac{h}{3} \left[ \underline{B} + 4 \underline{\Psi}^T \underline{B} \underline{\Psi} + 2 (\underline{\Psi}^T)^2 \underline{B} \underline{\Psi}^2 + 4 (\underline{\Psi}^T)^3 \underline{B} \underline{\Psi}^3 + \dots \right] \right)$$

where  $\underline{\Psi} = e^{-\underline{A}h}$

This approach requires the user to furnish two parameters  $T$  and  $h$ , for which he may have little feeling or concern. But the typical value of  $h$  that was found satisfactory to use in the algorithm was

$$h = \frac{1}{20 |\lambda_{\text{dom}}(\underline{A})|}$$

where  $\lambda_{\text{dom}}(\underline{A})$  is the dominant eigenvalue of  $\underline{A}$ , i.e.,

$$\lambda_i(\underline{A}) \leq \lambda_{\text{dom}}(\underline{A}) \quad i = 1, 2, \dots, n$$

In this study one example problem is solved by hand calculation and computer solution of the same problem is compared with it; it is seen that with this value of  $h$  solution converged to at least 6 significant figure accuracy.

#### A.4 EXAMPLE SOLUTION OF ALGEBRAIC RICCATI EQUATION (ARE)

The following Algebraic Riccati Equation is considered

$$\underline{A}^T \underline{R} + \underline{R} \underline{A} - \underline{R} \underline{B} \underline{P}^{-1} \underline{B} \underline{R} + \underline{Q} = 0$$

with

$$\underline{\underline{A}} = \begin{bmatrix} 0 & 1 \\ -1 & -1 \end{bmatrix} \quad \underline{\underline{B}} = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad P = 2 \quad \underline{\underline{Q}} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

and the following result has been got from the computer in 1 minute 11 seconds

$$\underline{\underline{R}} = \begin{bmatrix} 1.420079 & 0.4494897 \\ 0.4494897 & 0.7924826 \end{bmatrix}$$

## APPENDIX B

NUMERICAL SOLUTION OF  
THE LIAPUNOV MATRIX EQUATION

## B.1 THE LIAPUNOV MATRIX EQUATION

The Linear Matrix Equation

$$\underline{\underline{A}} \underline{\underline{X}} + \underline{\underline{X}} \underline{\underline{A}}^T + \underline{\underline{B}} = \underline{\underline{0}} \quad \text{B1}$$

where  $\underline{\underline{A}}$  and  $\underline{\underline{B}}$  are given ( $n \times n$ ) matrices and  $\underline{\underline{B}}$  is symmetric, is known as the Liapunov matrix equation. Note that  $\underline{\underline{X}}$  must be symmetric as well.

A number of procedures have been proposed for solving the Liapunov Matrix Equation (B1). These include direct solution, iterative procedures, finite series approximations, and transformation to various canonical forms. In this study, the following iterative procedure is used.

## B.2 METHOD OF SOLUTION

Let  $q$  be a positive parameter, let  $\underline{\underline{I}}$  be an ( $n \times n$ ) identity matrix and let

$$\underline{\underline{U}} = (q \underline{\underline{I}} - \underline{\underline{A}})^{-1} \quad \text{B2}$$

$$\underline{\underline{V}} = \underline{\underline{U}} (q \underline{\underline{I}} + \underline{\underline{A}}) \quad \text{B3}$$

$$\underline{\underline{W}} = 2q(\underline{\underline{U}} \underline{\underline{B}} \underline{\underline{U}}^T) \quad \text{B4}$$

If all eigenvalues of  $\underline{A}$  have negative real parts (which means a stable system),

$$\underline{Y} = \sum_{i=1}^L \underline{V}^{i-1} \underline{W} (\underline{V}^{i-1})^T \quad \text{B5}$$

converges and is the solution of Eq.(B1). L may go to infinity for exact solution.

The rate of convergence can be improved by using the sequence of partial sums

$$\underline{Y}_V = \sum_{i=1}^{2^V} \underline{V}^{i-1} \underline{W} (\underline{V}^{i-1})^T \quad \text{B6}$$

which can be obtained recursively from

$$\underline{Y}_0 = \underline{W} \quad \text{B7}$$

$$\underline{Y}_{V+1} = \underline{Y}_V + \underline{V}^{2^V} \underline{Y}_V (\underline{V}^{2^V})^T \quad \text{B8}$$

Ten application of the Recursion formula B8 yield the first 1024 terms of the series solution B5.

## APPENDIX C

PROOF OF THE MIN-MAX  
THEOREM

## C.1 DEFINITIONS OF NOTATION

$U$  denotes an arbitrary set with elements  $u$

$V$  denotes a set from a metric space with elements  $v$

$J(\underline{u}, \underline{v})$  denotes a real valued mapping of  $U \times V$  into  $\mathbb{R}$

$\hat{V}$  denotes the set of probability measures over  $V$   
with elements  $\hat{v}$ .

Assuming  $J(\underline{u}, \underline{v})$  is measurable with respect to elements of  $\hat{V}$  then by  $J(\underline{u}, \hat{v})$  we denote

$$J(\underline{u}, \hat{v}) = \int_V J(\underline{u}, \underline{v}) d\hat{v} \quad \underline{u} \in U, \hat{v} \in \hat{V} \quad C1$$

THE MIN-MAX THEOREM Assume

H1  $J(\underline{u}, \underline{v})$  is continuous with respect to  $\underline{u}$ , and  $\underline{v}$  for each  $\underline{u} \in U$  and  $\underline{v} \in V$ .

H2  $V$  is compact

H3 For each  $\hat{v} \in \hat{V}$  there exists a unique element, denoted  $\underline{u}^0(\hat{v}) \in U$  such that

$$\min_{\underline{u} \in U} J(\underline{u}, \hat{v}) = J(\underline{u}^0(\hat{v}), \hat{v}) \quad C2$$

and  $\underline{u}^0(\hat{v})$  is a continuous mapping from  $\hat{V}$  into  $U$ .

Then there exists a saddle point  $\underline{u} \in U$  and  $\hat{v} \in \hat{V}$  for  $J(\underline{u}, \hat{v})$  over  $U \times \hat{V}$  and

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) = \max_{\hat{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \hat{v}) = J(\underline{u}, \hat{v}) \quad C3$$

PROOF : The proof will proceed in three steps. First, we show existence of a pair  $\underline{u} \in U$  and  $\hat{v} \in V$  which satisfy

$$\max_{\hat{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \hat{v}) = J(\underline{u}, \hat{v}) \quad C4$$

Next we show that the pair  $\underline{u}, \hat{v}$  are a saddle point for  $J(\underline{u}, \hat{v})$  which implies

$$\max_{\hat{v} \in V} \min_{\underline{u} \in U} J(\underline{u}, \hat{v}) = \min_{\underline{u} \in U} \max_{\hat{v} \in V} J(\underline{u}, \hat{v}) \quad C5$$

Finally, we must show

$$\min_{\underline{u} \in U} \max_{\underline{v} \in V} J(\underline{u}, \underline{v}) = \min_{\underline{u} \in U} \max_{\hat{v} \in V} J(\underline{u}, \hat{v}) \quad C6$$

We begin with the proof of C4. Define

$$\Psi(\hat{v}) = \min_{\underline{u} \in U} J(\underline{u}, \hat{v}) \quad C7$$

Then  $\Psi(\hat{v})$  is upper semicontinuous with respect to  $\hat{v}$ . The set  $\hat{V}$  is compact by H2 and by the following Lemma.

Lemma : Compactness of  $V$  implies that  $\hat{V}$  is a compact set. Hence  $\Psi(\hat{v})$  achieves a maximum over  $\hat{V}$  for some element which we denote by  $\hat{v}^*$  which satisfies

$$\Psi(\hat{v}^*) \geq \Psi(\hat{v}) \quad \hat{v} \in \hat{V} \quad C8$$

By C7,

$$\psi(\hat{v}^*) = \max_{\hat{v} \in \hat{V}} \min_{u \in U} J(u, \hat{v}) \quad \text{C9}$$

However, by H3 there exists an element  $u^0(\hat{v}^*) \in U$  such that

$$\psi(\hat{v}^*) = \min_{u \in U} J(u, \hat{v}^*) = J(u^0(\hat{v}^*), \hat{v}^*) \quad \text{C10}$$

$$\text{Setting } \underline{u}^* = u^0(\hat{v}^*) \quad \text{C11}$$

and combining C9 and C10 yields C4.

Next we turn to the proof of C5. By the definition of a saddle point, C5 is satisfied if existence of a saddle point for  $J(\underline{u}, \hat{v})$  in  $U \times \hat{V}$  can be shown. We will show that the pair  $\underline{u}^*$ ,  $\hat{v}^*$  shown to exist and satisfy C4 constitutes such a saddle point. Let

$$J^* = J(\underline{u}^*, \hat{v}^*) \quad \text{C12}$$

Then we must show that

$$J(\underline{u}^*, \hat{v}) \leq J^* \leq J(\underline{u}, \hat{v}^*) \quad \forall \underline{u} \in U \text{ and } \hat{v} \in \hat{V} \quad \text{C13}$$

By the definition C1 of  $J(\underline{u}, \hat{v})$  we have that  $J(\underline{u}, \hat{v})$  is linear and hence concave with respect to  $\hat{v}$  for all  $\underline{u} \in U$ . Concavity of  $J(\underline{u}, \hat{v})$  over the convex set  $\hat{V}$  implies that for any  $0 \leq \xi \leq 1$  all  $\underline{u} \in U$  and all  $\hat{v} \in \hat{V}$ .

$$J(\underline{u}, (1-\xi)\hat{v}^* + \xi\hat{v}) \geq (1-\xi)\psi(\hat{v}^*) + \xi J(\underline{u}, \hat{v}) \quad \text{C14}$$

Using C7 and C14 yields

$$J(\underline{u}, (1-\xi)\hat{v}^* + \xi\hat{v}) \geq (1-\xi)\psi(\hat{v}^*) + \xi J(\underline{u}, \hat{v}) \quad \text{C15}$$

Replace  $\underline{u}$  in C15 by

$$\underline{u}_\xi^0 = u^0 \left( (1-\xi)\underline{v}^* + \xi\underline{v} \right) \quad \text{C16}$$

to obtain

$$\Psi \left( (1-\xi)\underline{v} + \xi\underline{v} \right) \geq (1-\xi) \Psi(\underline{v}^*) + \xi J(\underline{u}_\xi^0, \underline{v}) \quad \text{C17}$$

From C12 and C8

$$\Psi(\underline{v}^*) = J^* \geq \Psi \left( (1-\xi)\underline{v}^* + \xi\underline{v} \right) \quad \text{C18}$$

and therefore combining C18 with C17

$$J^* \geq (1-\xi)J + \xi J(\underline{u}_\xi^0, \underline{v}) \quad \text{C19}$$

or for  $0 \leq \xi \leq 1$

$$\xi J(\underline{u}_\xi^0, \underline{v}) \leq \xi J^* \quad \text{C20}$$

By continuity, setting  $\xi = 0$  in C20 and using C11 yields the left inequality of C13. Setting  $\xi = 0$  in C15 and noting C18 yields the right inequality of C13. This completes the proof of C13 which shows that  $\underline{u}$ ,  $\underline{v}$  is a saddle point and hence C5 holds.

To prove C16 we simply note that this corresponds to the equivalence lemma. The proof of this lemma also exists in the work of Howard Blum (R1).

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```

803 CONTINUE
DO 788 KCEK=1,NN
802 CONTINUE
DO 888 JCEK=1,NN
801 CONTINUE
DO 999 ICEK=1,NN
666 IF(IC.EQ.2)GO TO 552
II=1
IF(NPARAM.EQ.II)GO TO 901
II=II+1
IF(NPARAM.EQ.II)GO TO 902
II=II+1
IF(NPARAM.EQ.II)GO TO 903
II=II+1
IF(NPARAM.EQ.II)GO TO 904
DO 577 MM=1,NN
904 CONTINUE
DO 677 L=1,NN
903 CONTINUE
DO 777 K=1,NN
902 CONTINUE
DO 877 J=1,NN
901 CONTINUE
DO 977 I=1,NN
II=1
IF(NPARAM.NE.II)GO TO 401
IF(I.EQ.ICEK)GO TO 556
401 II=II+1
IF(NPARAM.NE.II)GO TO 402
IF((I.EQ.ICEK).AND.(J.EQ.JCEK))GO TO 557
402 II=II+1
IF(NPARAM.NE.II)GO TO 403
IF((I.EQ.ICEK).AND.(J.EQ.JCEK).AND.(K.EQ.KCEK))GO TO 558
403 II=II+1
IF(NPARAM.NE.II)GO TO 404
IF((I.EQ.ICEK).AND.(J.EQ.JCEK).AND.(K.EQ.KCEK).AND.(L.EQ.LC
1 TO 559
404 II=II+1
IF(NPARAM.NE.II)GO TO 977
IF((I.EQ.ICEK).AND.(J.EQ.JCEK).AND.(K.EQ.KCEK).AND.(L.EQ.LC
1.(MM.EQ.MCEK))GO TO 560
977 CONTINUE
877 CONTINUE
777 CONTINUE
677 CONTINUE
577 CONTINUE
560 V5=V5IN+(MM-1)*DELV5
559 V4=V4IN+(L-1)*DELV4
558 V3=V3IN+(K-1)*DELV3
557 V2=V2IN+(J-1)*DELV2
556 V1=V1IN+(I-1)*DELV1
C SYSTEM DESCRIPTION-- A, AND B MATRICES
A(1,1)=0. Δ
A(2,1)=-V1**2
A(2,2)=-V1
A(1,2)=1.
B(1,1)=0.
B(2,1)=1.
CALL EIGVAL(A,BB,C,N,POL,EIGEN)
DO 201 J=1,N
IF(REAL(EIGEN(J)).GT.0.)GO TO 58

```

```

IF(NPARAM.EQ.II)GO TO 987
V1=V1IN
I=1
V2=V2+DELV2
J=J+1
IF(J.LE.NN)GO TO 85
II=II+1
IF(NPARAM.EQ.II)GO TO 887
V1=V1IN
V2=V2IN
J=1
I=1
V3=V3+DELV3
K=K+1
IF(K.LE.NN)GO TO 85
II=II+1
IF(NPARAM.EQ.II)GO TO 787
V3=V3IN
V2=V2IN
V1=V1IN
K=1
J=1
I=1
V4=V4+DELV4
L=L+1
IF(L.LE.NN)GO TO 85
II=II+1
IF(NPARAM.EQ.II)GO TO 687
V4=V4IN
V3=V3IN
V2=V2IN
V1=V1IN
L=1
K=1
J=1
I=1
V5=V5+DELV5
MM=MM+1
IF(MM.LE.NN)GO TO 85
IF(ICEK.EQ.1.AND.JCEK.EQ.1.AND.KCEK.EQ.1.AND.LCEK.EQ.1.AND
1.1)XMINMA=PEROLD
IF(XMINMA.LT.PEROLD)GO TO 988
XMINMA=PEROLD
V1=V1IN+(ICEK-1)*DELV1
V2=V2IN+(JCEK-1)*DELV2
V3=V3IN+(KCEK-1)*DELV3
V4=V4IN+(LCEK-1)*DELV4
V5=V5IN+(MCEK-1)*DELV5
PRINT*, COORDINATES OF SUBOPTIMAL GRID POINT,
PRINT*, ICEK, JCEK, KCEK, LCEK, MCEK
PRINT*, VALUES OF THE PARAMETERS FOR SUBOPTIMAL GRID POINT
PRINT*, V1, V2, V3, V4, V5
GO TO 988
687 IF(ICEK.EQ.1.AND.JCEK.EQ.1.AND.KCEK.EQ.1.AND.LCEK.EQ.1)XMI
1OLD
IF(XMINMA.LT.PEROLD)GO TO 988
XMINMA=PEROLD
V1=V1IN+(ICEK-1)*DELV1
V2=V2IN+(JCEK-1)*DELV2
V3=V3IN+(KCEK-1)*DELV3
V4=V4IN+(LCEK-1)*DELV4

```

```

PRINT*, COORDINATES OF SUBOPTIMAL GRID POINT,
PRINT*, ICEK, JCEK, KCEK, LCEK
PRINT*, VALUES OF THE PARAMETERS FOR SUBOPTIMAL GRID POINT,
PRINT*, V1, V2, V3, V4
GO TO 988
787 IF(ICEK.EQ.1.AND.JCEK.EQ.1.AND.KCEK.EQ.1)XMINMA=PEROLD
PRINT*, MINMAX=, XMINMA
IF(XMINMA.LT.PEROLD)GO TO 988
XMINMA=PEROLD
V1=V1IN+(ICEK-1)*DELV1
V2=V2IN+(JCEK-1)*DELV2
V3=V3IN+(KCEK-1)*DELV3
PRINT*, COORDINATES OF SUBOPTIMAL GRID POINT,
PRINT*, ICEK, JCEK, KCEK
PRINT*, VALUES OF THE PARAMETERS FOR SUBOPTIMAL GRID POINT,
PRINT*, V1, V2, V3
GO TO 988
887 IF(ICEK.EQ.1.AND.JCEK.EQ.1)XMINMA=PEROLD
IF(XMINMA.LT.PEROLD)GO TO 988
XMINMA=PEROLD
V1=V1IN+(ICEK-1)*DELV1
V2=V2IN+(JCEK-1)*DELV2
PRINT*, COORDINATES OF SUBOPTIMAL GRID POINT,
PRINT*, ICEK, JCEK
PRINT*, VALUES OF THE PARAMETERS FOR SUBOPTIMAL GRID POINT,
PRINT*, V1, V2
GO TO 988
987 IF(ICEK.EQ.1)XMINMA=PEROLD
IF(XMINMA.LT.PEROLD)GO TO 988
XMINMA=PEROLD
V1=V1IN+(ICEK-1)*DELV1
PRINT*, COORDINATES OF SUBOPTIMAL GRID POINT,
PRINT*, ICEK
PRINT*, VALUES OF THE PARAMETERS FOR SUBOPTIMAL GRID POINT,
PRINT*, V1
988 IC=1
999 CONTINUE
888 CONTINUE
788 CONTINUE
688 CONTINUE
588 CONTINUE
END

```

```

SUBROUTINE RICATI(A,B,P,Q,N,M,H,R,GAIN,BARB)
DIMENSION A(10,10),B(10,10),P(10,10),Q(10,10),EXPA(10,10),
1),V(10),JC(10),BSQ(10,10),ASQ(10,10),QSQ(10,10),RB(10,10),
1,10),AT(10,10),ERR(10,10),ET(10,10),R(10,10),ARB(10,10),
1ARBY(10,10),EXPAT(10,10),BARB(10,10),GAIN(10,10)
V(1)=1.
CALL GJR(P,10,10,M,M,$50,JC,V)
DO 181 I=1,N
DO 181 J=1,M
181 BARB(I,J)=B(I,J)
CALL MXMLT(B,P,ARB,N,M,M,10,10)

```

.MAIN

17/27/79-16:03(,0)

```

1. DIMENSTON A(10,10),B(10,10),P(10,10),Q(10,10),R(10,10),ETC
2. 1,ABSEIG(10),GAIN(10,10),XIN(10),BARB(10,10),ARB(10,10)
3. COMPLEX POL,EIGEN
4. READ(5,7)N,M,NN,NPARAM
5. DATA V1IN,V2IN,V1FIN,V2FIN/1.0,1.0,2.0,2.0/
6. IST=0
7. P(1,1)=1.
8. Q(1,1)=1.
9. Q(1,2)=0.
0. Q(2,1)=0.
1. Q(2,2)=1.
2. XIN(1)=1.
3. XIN(2)=0.
4. 7 FORMAT(4I5)
5. NNM1=NN-1
6. II=1
7. IF(NPARAM.EQ.II)GO TO 701
8. II=II+1
9. IF(NPARAM.EQ.II)GO TO 702
0. II=II+1
1. IF(NPARAM.EQ.II)GO TO 703
2. II=II+1
3. IF(NPARAM.EQ.II)GO TO 704
4. DELV5=(V5FIN-V5IN)/NNM1
5. 704 DELV4=(V4FIN-V4IN)/NNM1
6. 703 DELV3=(V3FIN-V3IN)/NNM1
7. 702 DELV2=(V2FIN-V2IN)/NNM1
8. 701 DELV1=(V1FIN-V1IN)/NNM1
9. IC=1
0. II=1
1. IF(NPARAM.EQ.II)GO TO 801
2. II=II+1
3. IF(NPARAM.EQ.II)GO TO 802
4. II=II+1
5. IF(NPARAM.EQ.II)GO TO 803
6. II=II+1
7. IF(NPARAM.EQ.II)GO TO 804
8. DO 588 MCEK=1,NN
9. 804 CONTINUE
0. DO 688 LCEK=1,NN

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```

803 CONTINUE
DO 788 KCEK=1,NN
802 CONTINUE
DO 888 JCEK=1,NN
801 CONTINUE
DO 999 ICEK=1,NN
II=1
IF(NPARAM,EQ.II)GO TO 901
II=II+1
IF(NPARAM,EQ.II)GO TO 902
II=II+1
IF(NPARAM,EQ.II)GO TO 903
II=II+1
IF(NPARAM,EQ.II)GO TO 904
V5=V5IN+(MCEK-1)*DELV5
904 V4=V4IN+(LCEK-1)*DELV4
903 V3=V3IN+(KCEK-1)*DELV3
902 V2=V2IN+(JCEK-1)*DELV2
901 V1=V1IN+(ICEK-1)*DELV1
C SYSTEM DESCRIPTION-- A, AND B MATRICES
A(1,1)=0.
A(1,2)=1.
A(2,1)=-V1**2
A(2,2)=-V1
B(1,1)=0.
B(2,1)=1.
CALL EIGVAL(A,BB,C,N,POL,EIGEN)
DO 201 J=1,N
IF(REAL(EIGEN(J)).GT.0.)GO TO 58
201 CONTINUE
GO TO 59
58 PRINT*,,SYSTEM IS UNSTABLE FOR THESE PARAMETERS WITHOUT FEE
IST=1
59 CONTINUE
DO 55 I=1,N
55 ABSEIG(I)=CABS(EIGEN(I))
BIGEIG=ABSEIG(1)
DO 56 I=2,N
56 BIGEIG=AMAX1(ABSEIG(I),BIGEIG)
H=1./(20.*BIGEIG)
CALL RICATI(A,B,P,Q,N,M,H,R,GAIN,BARB)
IF(IST.EQ.1)GO TO 16
GO TO 69
16 CALL MXMLT(BARB,GAIN,ARB,N,M,N,10,10)
CALL MXADD(A,ARB,A,N,N,10)
CALL EIGVAL(A,BB,C,N,POL,EIGEN)
IST=0
DO 17 I=1,N
IF(REAL(EIGEN(I)).GT.0.)GO TO 68
17 CONTINUE
GO TO 69
68 PRINT*,,SYSTEM UNSTABLE FOR BELOW PARAMETERS WITH STATE FEE
PRINT*,,PARAMETER 1--PARAMETER 2--PARAMETER 3--PARAMETER
PRINT*,V1
69 CONTINUE
CALL OPTIMA(XIN,R,N,M,OPT)
PRINT*,,OPTIMUM VALUE=,,OPT
II=1
IF(NPARAM,EQ.II)GO TO 987
II=II+1
IF(NPARAM,EQ.II)GO TO 887

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II=II+1
IF(NPARAM,EQ,II)GO TO 787
II=II+1
IF(NPARAM,EQ,II)GO TO 687
IF(ICEK,EQ,1.AND,JCEK,EQ,1.AND,KCEK,EQ,1.AND,LCEK,EQ,1.AND,M
1.1)GO TO 500
GO TO 600
687 IF(ICEK,EQ,1.AND,JCEK,EQ,1.AND,KCEK,EQ,1.AND,LCEK,EQ,1)GO TO
GO TO 600
787 IF(ICEK,EQ,1.AND,JCEK,EQ,1.AND,KCEK,EQ,1)GO TO 500
GO TO 600
887 IF(ICEK,EQ,1.AND,JCEK,EQ,1)GO TO 500
GO TO 600
987 IF(ICEK,EQ,1)GO TO 500
600 CONTINUE
IF(OPT,LT,OPTOLD)GO TO 999
IF(OPT,EQ,OPTOLD)GO TO 988
500 OPTOLD=OPT
DO 301 I=1,M
DO 301 J=1,N
301 OPGAIN(I,J)=GAIN(I,J)
II=1
IF(NPARAM,EQ,II)GO TO 501
II=II+1
IF(NPARAM,EQ,II)GO TO 502
II=II+1
IF(NPARAM,EQ,II)GO TO 503
II=II+1
IF(NPARAM,EQ,II)GO TO 504
V5OPT=V5
504 V4OPT=V4
503 V3OPT=V3
502 V2OPT=V2
501 V1OPT=V1
PRINT*,,MAXIMIN SOL-OPTOLD=,,OPTOLD
PRINT*,,GAIN VECTOR FOR MAXMIN SOL,
PRINT*,((GAIN(I,J),J=1,N),I=1,M)
PRINT*,,PARAMETERS FOR MAXMIN SOL,,v1
GO TO 999
988 PRINT*,,THERE EXISTS ANOTHER MAXMIN SOL,,v1,v2
999 CONTINUE
888 CONTINUE
788 CONTINUE
688 CONTINUE
588 CONTINUE
II=1
IF(NPARAM,EQ,II)GO TO 601
II=II+1
IF(NPARAM,EQ,II)GO TO 602
II=II+1
IF(NPARAM,EQ,II)GO TO 603
II=II+1
IF(NPARAM,EQ,II)GO TO 604
DO 188 MK=1,NN
604 CONTINUE
DO 288 LK=1,NN
603 CONTINUE
DO 388 KK=1,NN
602 CONTINUE
DO 488 JK=1,NN
601 CONTINUE

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DO 499 IK=1,NN
  II=1
  IF(NPARAM.EQ.II)GO TO 101
  II=II+1
  IF(NPARAM.EQ.II)GO TO 102
  II=II+1
  IF(NPARAM.EQ.II)GO TO 103
  II=II+1
  IF(NPARAM.EQ.II)GO TO 104
  V5=V5IN+(MK-1)*DELV5
104 V4=V4IN+(LK-1)*DELV4
103 V3=V3IN+(KK-1)*DELV3
102 V2=V2IN+(JK-1)*DELV2
101 V1=V1IN+(IK-1)*DELV1
C SYSTEM DESCRIPTION-- A,AND B MATRICES
  A(1,1)=0.
  A(1,2)=1.
  A(2,1)=-V1**2
  A(2,2)=-V1
  B(1,1)=0.
  B(2,1)=1.
  CALL PERFOR(A,B,OPGAIN,P,Q,XIN,N,M,ZOR)
  II=1
  IF(NPARAM.EQ.II)GO TO 487
  II=II+1
  IF(NPARAM.EQ.II)GO TO 387
  II=II+1
  II=II+1
  IF(NPARAM.EQ.II)GO TO 287
  IF(NPARAM.EQ.II)GO TO 187
  IF(IK.EQ.1.AND.JK.EQ.1.AND.KK.EQ.1.AND.LK.EQ.1.AND.MK.EQ.1)
1400
  GO TO 399
187 IF(IK.EQ.1.AND.JK.EQ.1.AND.KK.EQ.1.AND.LK.EQ.1)GO TO 400
  GO TO 399
287 IF(IK.EQ.1.AND.JK.EQ.1.AND.KK.EQ.1)GO TO 400
  GO TO 399
387 IF(IK.EQ.1.AND.JK.EQ.1)GO TO 400
  GO TO 399
487 IF(IK.EQ.1)GO TO 400
399 CONTINUE
  PRINT*,',ZOR VALUE=',ZOR
  IF(ZOR.LE.ZOROLD)GO TO 499
400 ZOROLD=ZOR
  II=1
  IF(NPARAM.EQ.II)GO TO 202
  II=II+1
  IF(NPARAM.EQ.II)GO TO 203
  II=II+1
  IF(NPARAM.EQ.II)GO TO 204
  II=II+1
  IF(NPARAM.EQ.II)GO TO 205
  V5ZOR=V5
205 V4ZOR=V4
204 V3ZOR=V3
203 V2ZOR=V2
202 V1ZOR=V1
  PRINT*,',ZOROLD=',ZOROLD,',V1ZOR=',V1
499 CONTINUE
488 CONTINUE
388 CONTINUE

```

288 CONTINUE

188 CONTINUE

IF(OPTOLD.EQ.ZOROLD)GO TO 215  
PRINT\*,SADDLE POINT DOES NOT EXIST,  
GO TO 216

215 PRINT\*,SADDLE POINT EXISTS,  
PRINT\*,V1OPT

216 STOP

END

SUBROUTINE RICATI(A,B,P,Q,N,M,H,R,GAIN,BARB)  
DIMENSION A(10,10),B(10,10),P(10,10),Q(10,10),EXPA(10,10),W  
1),V(10),JC(10),BSQ(10,10),ASQ(10,10),QSQ(10,10),RB(10,10),R  
1,10),AT(10,10),ERR(10,10),ET(10,10),R(10,10),ARB(10,10),  
1ARBY(10,10),EXPAT(10,10),BARB(10,10),GAIN(10,10)

V(1)=1.

CALL GJR(P,10,10,M,M,\$50,JC,V)

DO 181 I=1,N

DO 181 J=1,M

181 BARB(I,J)=B(I,J)

CALL MXMLT(B,P,ARB,N,M,M,10,10)

CALL MXTRN(B,B,N,M,10,10)

CALL MXMLT(ARB,B,BSQ,N,M,N,10,10)

DO 88 I=1,N

DO 88 J=1,N

B(I,J)=BSQ(I,J)

88 W(I,J)=BSQ(I,J)

CALL MXSCA(A,N,N,10,-1.)

CALL MATEXP(A,EXPA,N,H)

CALL MXMLT(B,EXPA,ARB,N,N,N,10,10)

CALL MXTRN(EXPA,EXPAT,N,N,10,10)

CALL MXMLT(EXPAT,ARB,ARBY,N,N,N,10,10)

CALL MXSCA(ARBY,N,N,10,4.)

CALL MXADD(B,ARBY,W,N,N,10)

CALL MXSCA(ARBY,N,N,10,0.5)

CALL MXMLT(EXPAT,ARBY,ARB,N,N,N,10,10)

CALL MXMLT(ARB,EXPA,ARBY,N,N,N,10,10)

CALL MXADD(W,ARBY,W,N,N,10)

CALL MXSCA(ARBY,N,N,10,2.)

CALL MXMLT(EXPAT,ARBY,ARB,N,N,N,10,10)

CALL MXMLT(ARB,EXPA,ARBY,N,N,N,10,10)

CALL MXADD(W,ARBY,W,N,N,10)

S=H/3.

CALL MXSCA(W,N,N,10,S)

V(1)=1.

CALL GJR(W,10,10,N,N,\$50,JC,V)

GO TO 152

155 CONTINUE

DO 183 I=1,N

DO 183 J=1,N

183 W(I,J)=0.

DO 302 I=1,N

302 W(I,I)=1.

152 CALL MXMLT(W,BSQ,R,N,N,N,10,10)

```

DO 99 I=1,N
DO 99 J=1,N
99 ERR(I,J)=10.E-05
CALL MXSCA(A,N,N,10,-1.)
111 CALL MXMLT(R,BSQ,ARB,N,N,N,10,10)
CALL MXSUB(A,ARB,ASQ,N,N,10)
55 CALL MXMLT(RB,R,ARB,N,N,N,10,10)
CALL MXADD(Q,ARB,QSQ,N,N,10)
DO 98 I=1,N
DO 98 J=1,N
98 ROLD(I,J)=R(I,J)
CALL LIAPUN(ASQ,QSQ,N,R)
DO 78 I=1,N
DO 77 J=1,N
IF(ABS(R(I,J)-ROLD(I,J)).LE.ERR(I,J))GO TO 77
GO TO 111
77 CONTINUE
78 CONTINUE
CALL MXMLT(A,R,ET,N,N,N,10,10)
CALL MXTRN(A,AT,N,N,10,10)
CALL MXMLT(R,AT,ARB,N,N,N,10,10)
CALL MXADD(ET,ARB,ET,N,N,10)
CALL MXMLT(R,BSQ,ARB,N,N,N,10,10)
CALL MXMLT(ARB,R,ARBY,N,N,N,10,10)
CALL MXSUB(ET,ARBY,ET,N,N,10)
CALL MXADD(ET,Q,ET,N,N,10)
DO 73 I=1,N
DO 72 J=1,N
IF(ABS(ET(I,J)).LE.ERR(I,J))GO TO 72
GO TO 111
72 CONTINUE
73 CONTINUE
WRITE(6,64)
PRINT*,,(R(I,J),J=1,N),I=1,N)
GO TO 51
50 WRITE(6,140)
GO TO 155
51 CONTINUE
CALL MXTRN(BARB,BARB,N,M,10,10)
CALL MXMLT(BARB,R,ARB,M,N,N,10,10)
CALL MXMLT(P,ARB,GAIN,M,M,N,10,10)
CALL MXSCA(GAIN,M,N,10,-1.)
V(1)=1.
CALL GJR(P,10,10,M,M,$50,JC,V)
PRINT*,,GAIN VECTOR K IS GIVEN BELOW,
PRINT*,,(GAIN(I,J),J=1,N),I=1,M)
64 FORMAT(1X, 'RICATI MATRIX IS GIVEN BELOW, ,')
140 FORMAT(1X, 'INVERSE DOES NOT EXIST, ,')
RETURN
END

```

```

SUBROUTINE LIAPUN(ASQ,QSQ,N,R)
DIMENSION ASQ(10,10),QSQ(10,10),R(10,10),SIDEN(10,10),U(10,
1,VV(10,10),UT(10,10),Y(10,10),Z(10,10),ZT(10,10),ARB(10,10

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```

1, JC(10), ZEN(10,10)
S=0.1
DO 101 J=1,N
101 SIDEN(J,J)=S
CALL MXSUB(SIDEN, ASQ, U, N, N, 10)
V(1)=1.
CALL GJR(U, 10, 10, N, N, $50, JC, V)
GO TO 105
102 CONTINUE
DO 103 I=1,N
DO 103 J=1,N
103 U(I,J)=0.
DO 104 I=1,N
104 U(I,I)=1.
105 CONTINUE
CALL MXADD(SIDEN, ASQ, VV, N, N, 10)
CALL MXMLT(U, VV, ARB, N, N, N, 10, 10)
DO 181 I=1,N
DO 181 J=1,N
181 Z(I,J)=ARB(I,J)
CALL MXMLT(U, QSQ, ARB, N, N, N, 10, 10)
CALL MXTRN(U, UT, N, N, 10, 10)
CALL MXMLT(ARB, UT, Y, N, N, N, 10, 10)
S2=2.*S
CALL MXSCA(Y, N, N, 10, S2)
DO 203 KK=1,10
IF(KK.EQ.1)GO TO 190
CALL MXMLT(Z, Z, ARB, N, N, N, 10, 10)
DO 182 I=1,N
DO 182 J=1,N
182 Z(I,J)=ARB(I,J)
190 CALL MXMLT(Z, Y, ARB, N, N, N, 10, 10)
CALL MXTRN(Z, ZT, N, N, 10, 10)
CALL MXMLT(ARB, ZT, ZEN, N, N, N, 10, 10)
CALL MXADD(Y, ZEN, Y, N, N, 10)
203 CONTINUE
DO 204 I=1,N
DO 204 J=1,N
204 R(I,J)=Y(I,J)
GO TO 51
50 WRITE(6,140)
GO TO 102
51 CONTINUE
140 FORMAT(1X, 'INVERSE DOES NOT EXIST,')
RETURN
END

```

```

SUBROUTINE MATEXP(A, EXPA, N, T)
DIMENSION A(10,10), EXPA(10,10), ST(10,10), ENTEXP(10,10)
DO 1 I=1,N
DO 1 J=1,N
ST(I,J)=A(I,J)*T
1 EXPA(I,J)=ST(I,J)
IPOWER=100

```

```

DO 7 I=2, IPOWER
EPOWER=IPOWER-I+2
DO 5 J=1, N
DO 3 K=1, N
3 ENTEXP(J, K)=EXPA(J, K)/EPOWER
5 ENTEXP(J, J)=ENTEXP(J, J)+1.0
7 CALL MXMLT(ST, ENTEXP, EXPA, N, N, N, 15, 15)
DO 9 J=1, N
9 EXPA(J, J)=EXPA(J, J)+1.0
RETURN
END

```

```

SUBROUTINE PERFOR(A, B, GAIN, P, Q, XIN, N, M, PER)
DIMENSION A(10, 10), B(10, 10), P(10, 10), Q(10, 10), GAIN(10, 10), X
1, ARB(10, 10), ARBY(10, 10), ASQ(10, 10), QSQ(10, 10)
CALL MXMLT(B, GAIN, ARB, N, M, N, 10, 10)
CALL MXADD(A, ARB, ASQ, N, N, 10, 10)
CALL MXMLT(P, GAIN, ARB, M, M, N, 10, 10)
CALL MXTRN(GAIN, GAIN, M, N, 10, 10)
CALL MXMLT(GAIN, ARB, ARBY, N, M, N, 10, 10)
CALL MXADD(Q, ARBY, QSQ, N, N, 10, 10)
CALL LIAPUN(ASQ, QSQ, N, R)
CALL OPTIMA(XIN, R, N, M, PER)
RETURN
END

```

```

SUBROUTINE OPTIMA(XIN, R, N, M, OPT)
DIMENSION XIN(10), RR(10), R(10, 10)
CALL MXMLT(R, XIN, RR, N, N, 1, 10, 10)
CALL MXTRN(XIN, XIN, N, 1, 10, 10)
CALL MXMLT(XIN, RR, OPT, 1, N, 1, 10, 10)
RETURN
END

```

```

SUBROUTINE EIGVAL(A, BB, C, N, POL, EIGEN)
DIMENSION A(10, 10), BB(10, 10), C(10, 10), P(10), POL(10), EIGEN(1
COMPLEX POL, EIGEN
EPS=.000001
KMAX=50
DO 1 I=1, N

```

```

DO 1 J=1,M
1 BB(I,J)=A(I,J)
DO 2 M=1,N
P(M)=0.0
FM=M
DO 21 I=1,N
21 P(M)=P(M)+BB(I,I)/FM
DO 3 I=1,N
3 BB(I,I)=BB(I,I)-P(M)
CALL MXMLT(A, BB, C, N, N, N, 10, 10)
DO 6 K=1,N
DO 6 L=1,N
6 BB(K,L)=C(K,L)
2 CONTINUE
NI=N
DO 41 I=1,N
L=N-I+1
IF(P(L).NE.0.)GO TO 51
41 NI=N-I
51 CONTINUE
IF(NI.EQ.0)GO TO 22
DO 4 I=1,NI
J=NI-I+1
4 POL(J)=CMPLX(-P(I),0.0)
NP1=NI+1
POL(NP1)=CMPLX(1.0,0.0)
CALL ROOTCP(POL,NI, EPS, KMAX, EIGEN, J, $66)
GO TO 85
22 NP1=NI+1
85 CONTINUE
DO 23 I=NP1,N,1
23 EIGEN(I)=CMPLX(0.0,0.0)
WRITE(6,11)(EIGEN(I),I=1,N)
GO TO 31
66 WRITE(6,77)J
31 CONTINUE
11 FORMAT(1X,**EIGENVALUES ARE AS FOLLOWS**,///,1X,F12.6,5X,F
77 FORMAT(1X,I4,,TH ROOT DIVERGED,///)
RETURN
END

```

IBANK 5393 DBANK

MAIN

08/14/79 14:05:57

S	001000 042521	17234 IBANK WORDS DECIMAL
	043000 061020	7185 DBANK WORDS DECIMAL
RESS	035763	

SEGMENT \$MAIN\$ 001000 042521 043000 061020

(1) 002200 002134 \$(2) 043000 043002

\$(3) 002132 002146

\$(5) 002147 002147

\$(2) 043003 043314